



Synchronization and Control Methods for grid-connecting Renewable Energy Sources

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Outline

1. Introduction
2. Power electronics: The enabling technology for RES
3. Grid side converter controller
4. Synchronization methods
5. Conclusions



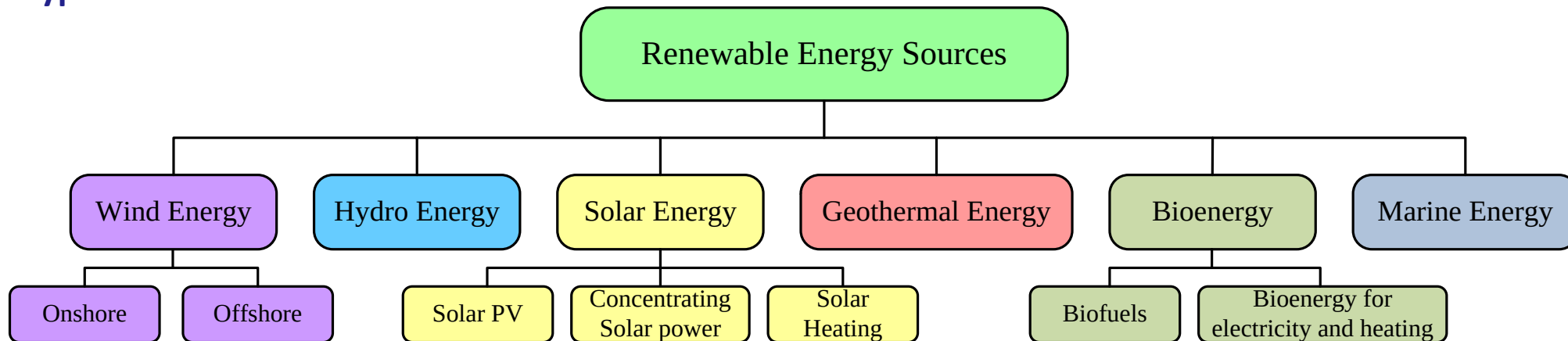
1. Introduction

Renewable Energy Sources: Definition – Current Status – Future Prospect – Challenging Aspects

Renewable Energy Sources (RES) - definition

- Renewable energy is generally defined as the energy from a source that is not depleted when used or the energy from resources which are naturally replenished on a human timescale such as sunlight, wind, rain, tides, waves and geothermal heat

Type of RES:





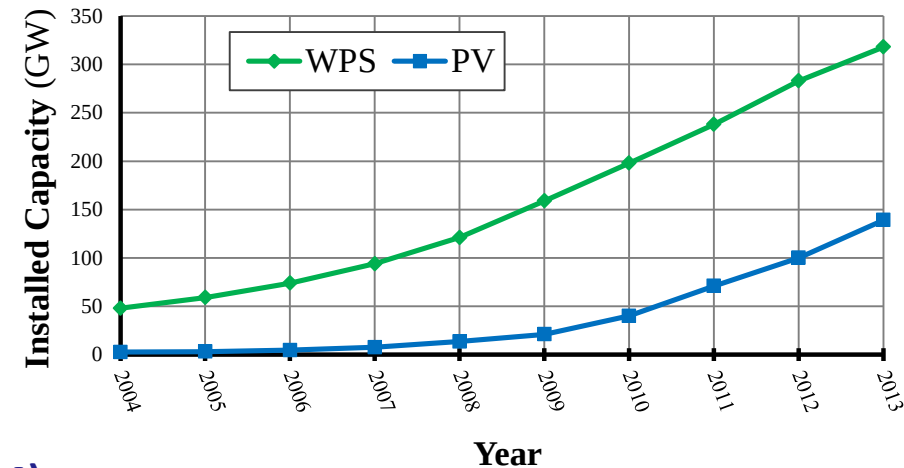
1. Introduction

Renewable Energy Sources: Definition – **Current Status** – Future Prospect – Challenging Aspects

RES are receiving considerable attention throughout the world

- Price of fossil fuels
- Environmental impact of fossil fuels / Climate change

**Worldwide installed capacity (GigaWatt)
for Wind Power Systems (WPS) and solar PV systems**



Share of energy consumption originating from RES (2014)

- European Union → 14%
- Cyprus → 7%



1. Introduction

Renewable Energy Sources: Definition – Current Status – **Future Prospect** – Challenging Aspects

European Union aims → “20-20-20 targets” by 2020

- **20% reduction of green house gas emissions** compared to 1990 levels
- **20%** reduction in primary energy use through improved **energy efficiency**
- **20%** share of **energy consumption** originating from **renewable energy sources (RES)**

For Cyprus the target is define at 13% of energy consumption to produce from RES by 2020

- Autonomous-isolated power grid → Stability problems

European Union 2030 Framework for Climate & Energy

- **40% reduction of green house gas emissions** compared to 1990 levels
- **27%** reduction in primary energy use through improved **energy efficiency**
- **27%** share of **energy consumption** originating from **renewable energy sources (RES)**



1. Introduction

Renewable Energy Sources: Definition – Current Status – Future Prospect – **Challenging Aspects**

The increasing penetration of RES:

- enables the **reduction of green house gas emissions** (environmental benefits)
- **can raise stability issues** regarding the operation of the power system
- **can affect the power quality** of the power system
- can lead to **cascading and blackout events** in the power system

Grid regulations for grid-connecting RES are evolving to ensure the proper power system operation

- RES should inject **high quality currents** even under highly harmonic distorted grid voltage
- RES should be **remain interconnected** under **balanced** or **unbalanced grid faults**
- RES should ensure a **fast and appropriate Fault Ride Through (FRT) operation** under grid faults
- RES should **provide frequency (active power) and voltage (reactive power) support** to the grid

1. Introduction

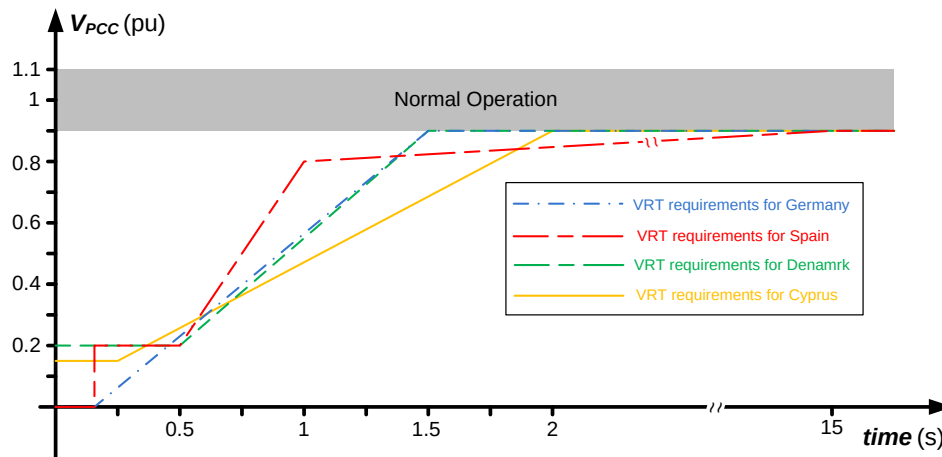
Renewable Energy Sources: Definition – Current Status – Future Prospect – Challenging Aspects

The design of RES should be advanced in order to achieve:

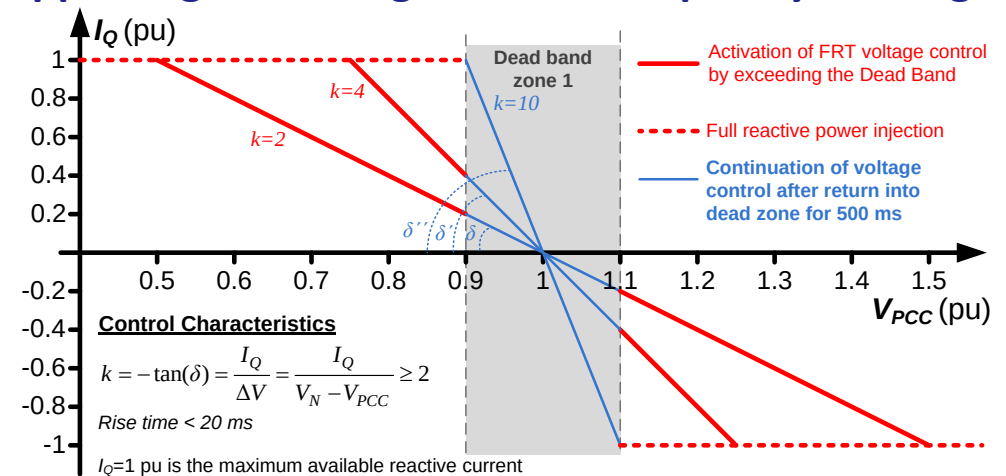
- high quality current injection
- immunity against voltage harmonic distortion
- fast and dynamic response
- proper FRT operation under abnormal grid conditions

Under a balanced or unbalanced grid faults

RES should keep its synchronization and remain grid-connected



RES should ensure a dynamic FRT operation for supporting the voltage and the frequency of the grid

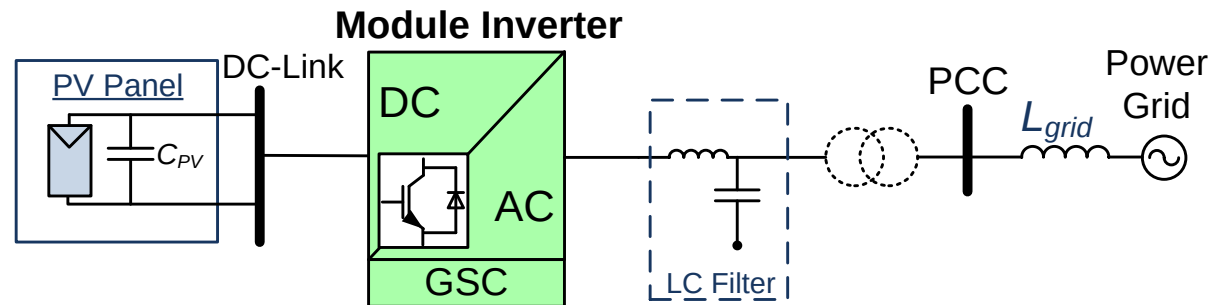


2. Power Electronics – The enabling technology for RES

Power Electronics: PV topologies – WPS topologies – Grid Side Converter (Inverter)

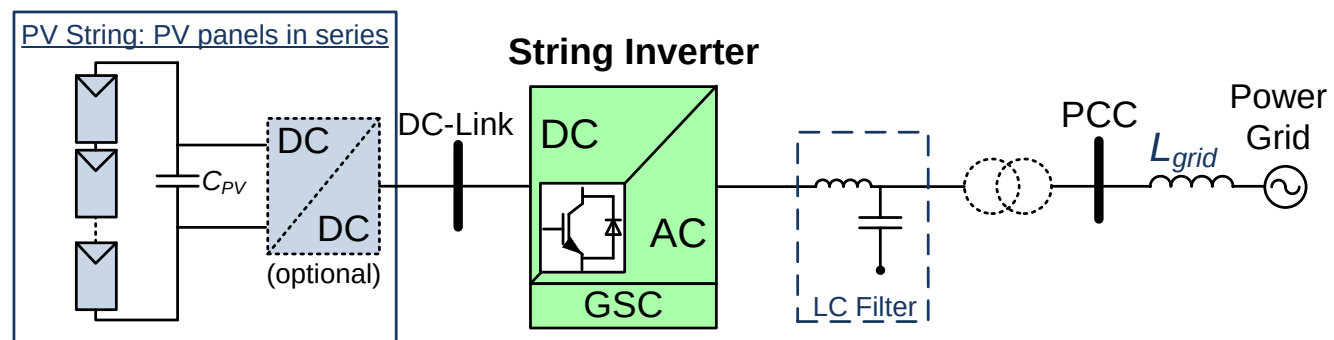
Grid-Connected PV system with Module Inverter

- Small systems
- ~ 300 W
- 1-phase Module Inverter



Grid-Connected PV system with String Inverter

- Residential Applications
- 1 – 10 kW
- 1- or 3-phase String Inverter

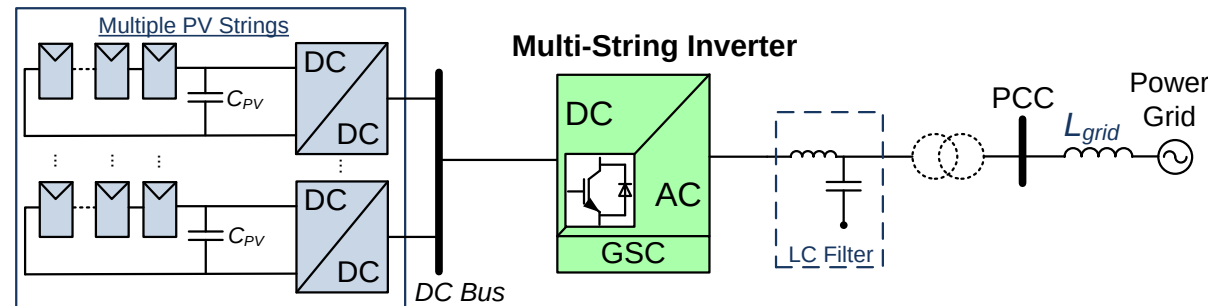


2. Power Electronics – The enabling technology for RES

Power Electronics: PV topologies – WPS topologies – Grid Side Converter (Inverter)

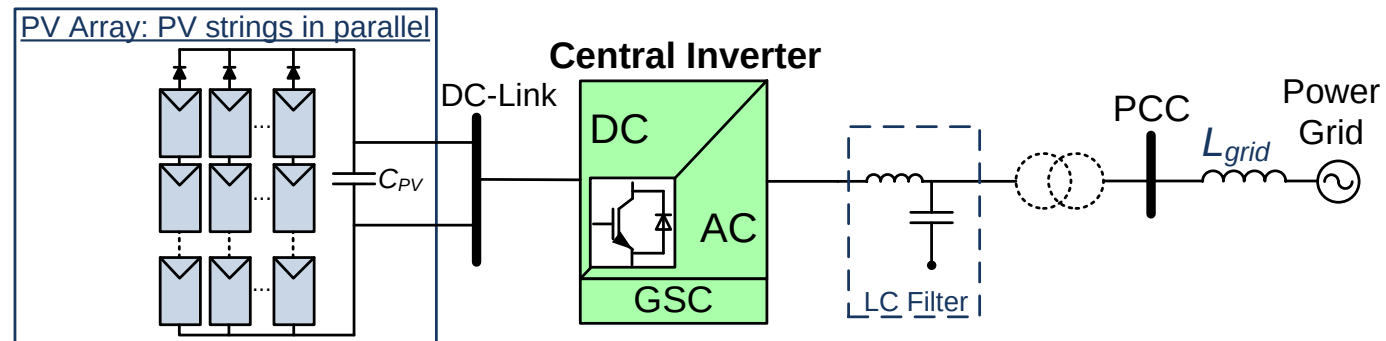
Grid-Connected PV system with Multi-String Inverter

- Residential/Commercial Applications
- 5 – 30 kW
- 1- or 3-phase Multi-String Inverter



Grid-Connected PV system with Central Inverter

- Commercial Applications
- > 30 kW
- 3-phase Central Inverter

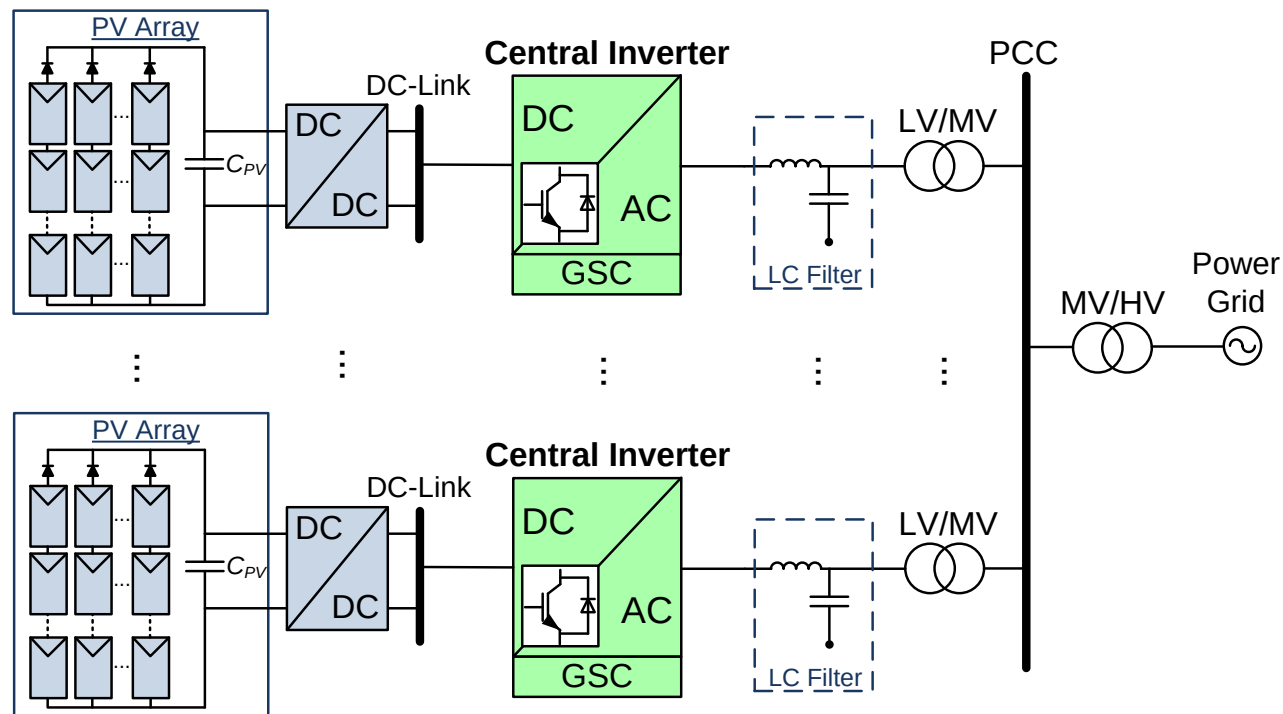


2. Power Electronics – The enabling technology for RES

Power Electronics: PV topologies – WPS topologies – Grid Side Converter (Inverter)

Grid-Connected PV system with Multiple PV arrays and Central Inverters

- Utility scale/Commercial Applications
- > 100 kW
- Multiple 3-phase Central Inverters

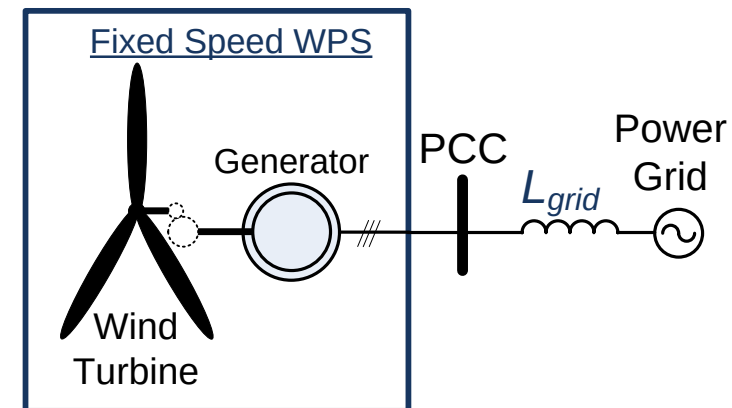


2. Power Electronics – The enabling technology for RES

Power Electronics: PV topologies – **WPS topologies** – Grid Side Converter (Inverter)

Direct Grid-Connected WPS system

- 1970-2000
- Fixed Speed → Low efficiency
- Generator Type: - usually Induction Generators
- synchronous generators have also be used
- Vulnerable to grid disturbances

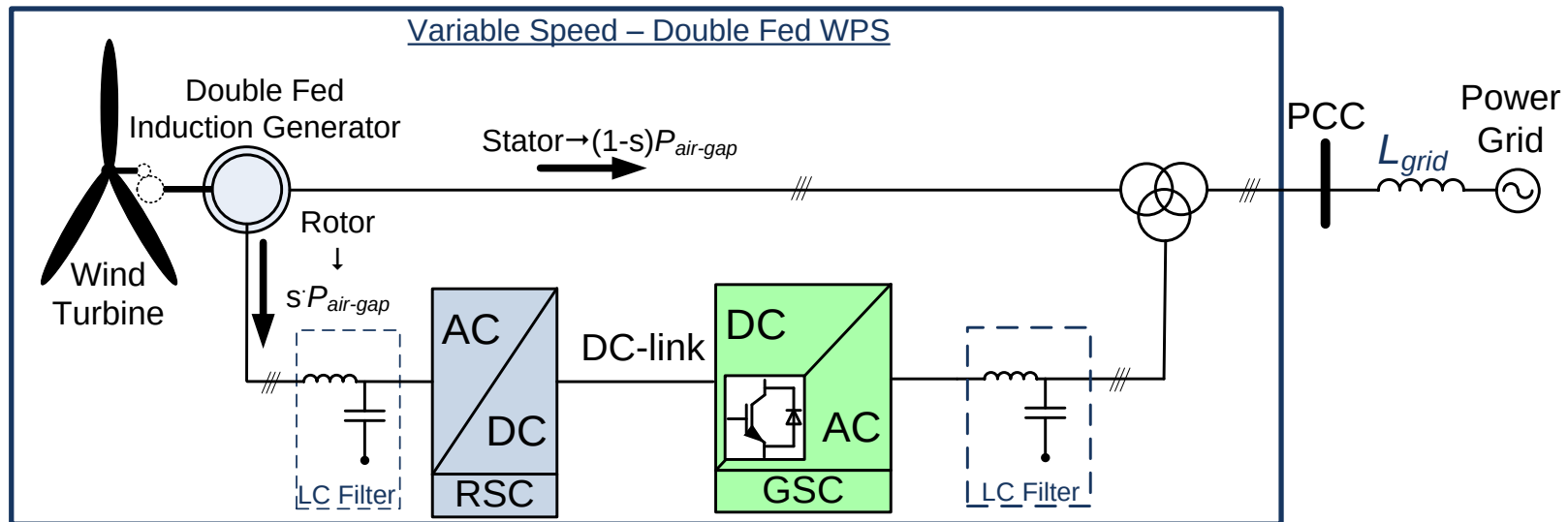


2. Power Electronics – The enabling technology for RES

Power Electronics: PV topologies – **WPS topologies** – Grid Side Converter (Inverter)

Double Fed WPS system

- 2000-today (most widely used solution)
- Variable Speed (within a range) → Increased efficiency
- Absorbs reactive power
- Power Electronics: ~ 30% of the rated power (Rotor Side)
- Generator Type: Wound Rotor Induction Generator (use of slip rings: increased maintenance)
- Vulnerable to grid disturbances (difficult to meet modern grid requirements)

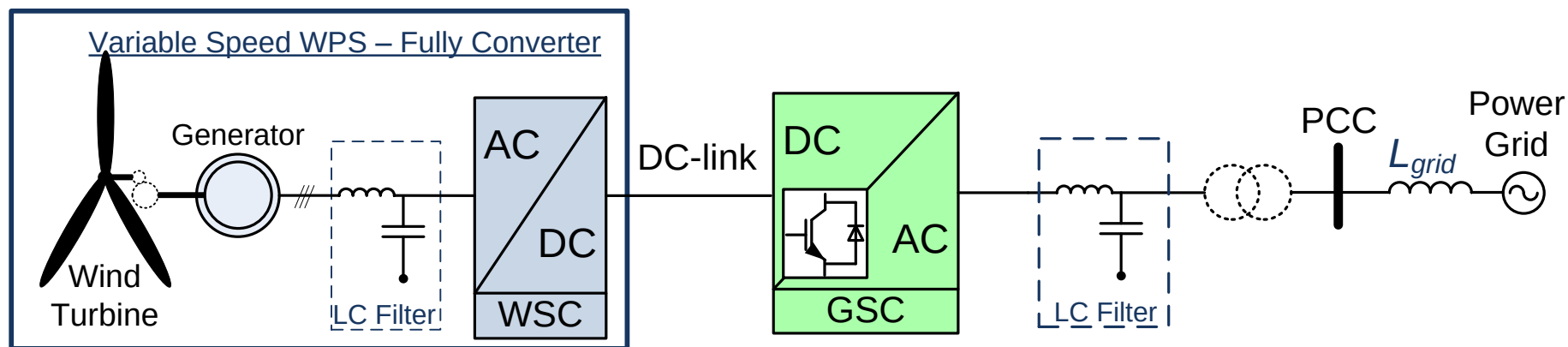


2. Power Electronics – The enabling technology for RES

Power Electronics: PV topologies – **WPS topologies** – Grid Side Converter (Inverter)

Fully Converter WPS system

- 2005-today (state-of-the-art)
- Variable Speed → High efficiency in any wind speed
- Power Electronics: 100% of the rated power (higher cost) (2 Back to Back bidirectional converters)
- Generator Type: Induction or Synchronous Generator
- Absorbs/Delivers reactive power (can provide voltage and frequency support to the power grid)
- Enables Fault Ride Through operation under grid faults (meet modern grid requirements)

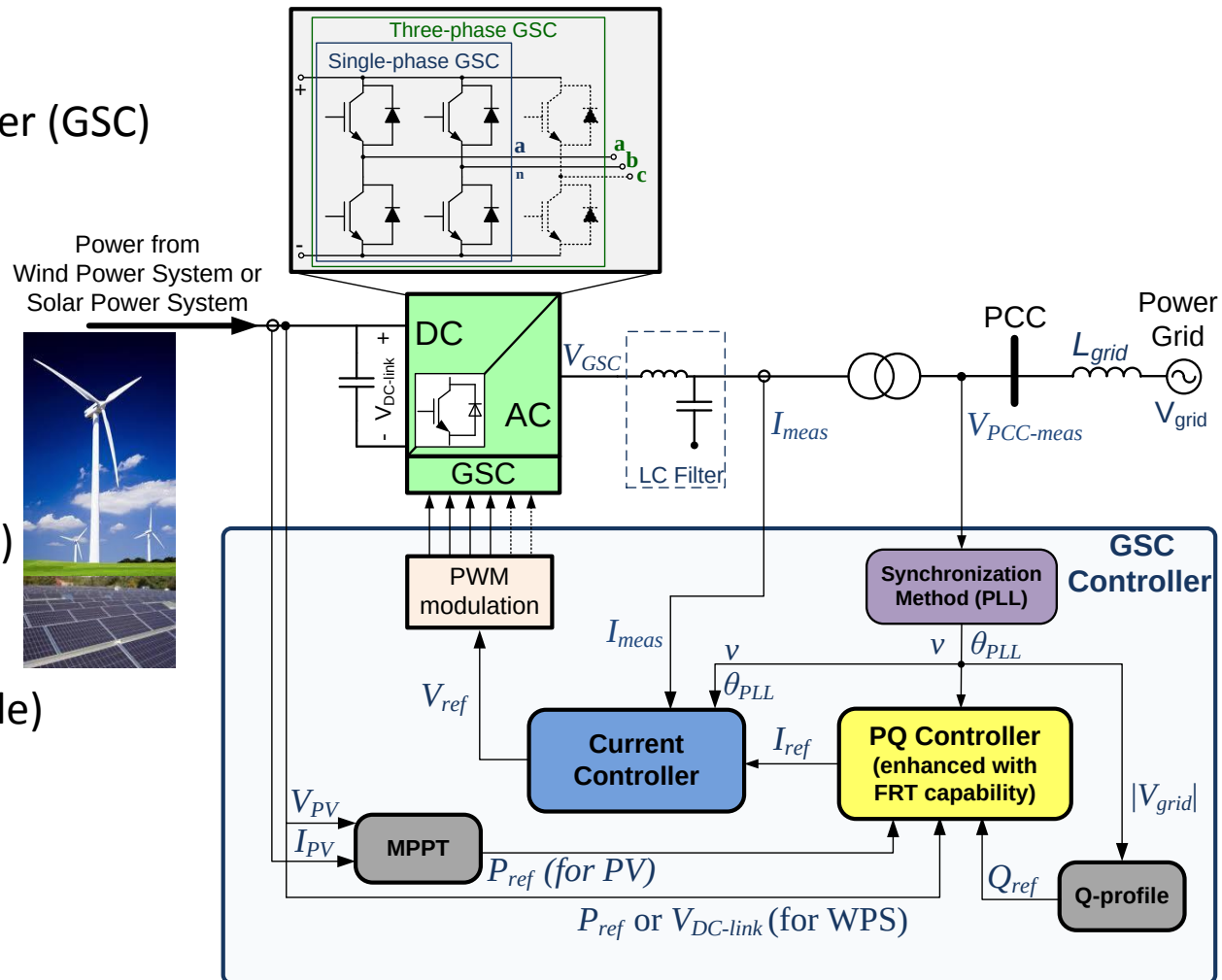


2. Power Electronics – The enabling technology for RES

Power Electronics: PV topologies – WPS topologies – Grid Side Converter (Inverter)

Similarity in PV and WPS topologies

- Grid Side power electronic Converter (GSC)
- Single- or three-phase inverter
- MOSFETs or IGBTs
- GSC Controller (Main modules)
 - V and I measurements
 - Park Transformations ($\alpha\beta$ and dq)
 - **Synchronization method**
 - PQ controller (MPPT and Q-profile)
 - Current controller
 - PWM modulations

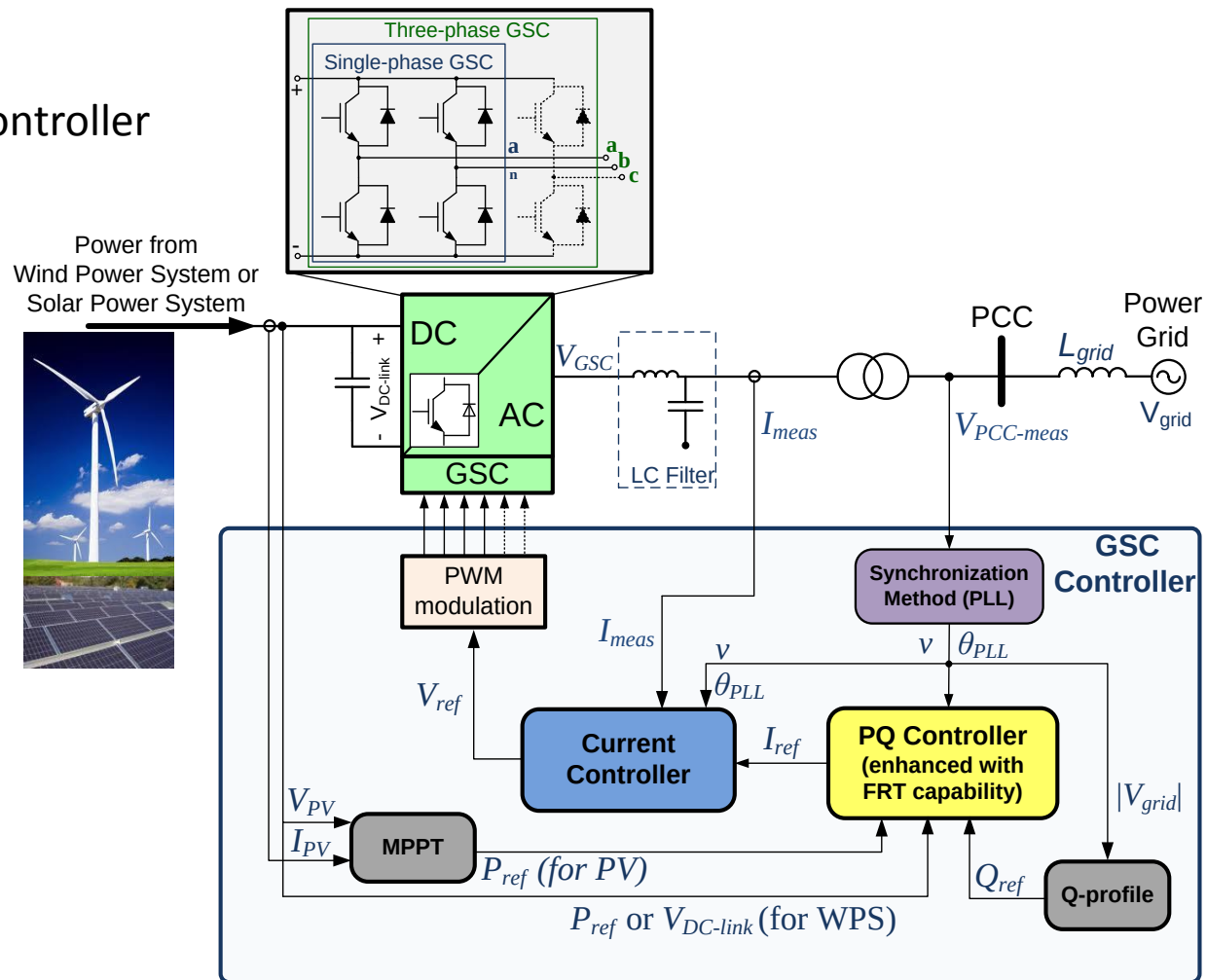


3. Grid Side Converter Controller

GSC Controller: PWM – Transformations – Synchronization – Current Controller – PQ controller

GSC controller

- Developed in an embedded microcontroller within the inverter
- Sampling frequency ~ 10 kHz (voltage and current sensors)
- Real-time control
- Ensure the proper operation of RES



3. Grid Side Converter Controller

GSC Controller: Transformations – Synchronization – PQ controller – Current Controller – PWM

Park Transformations

- a-b-c natural frame
- αβ-frame: stationary reference frame
- **dq-frame**: rotating frame with the synchronous rotational speed

abc-frame → αβ-frame

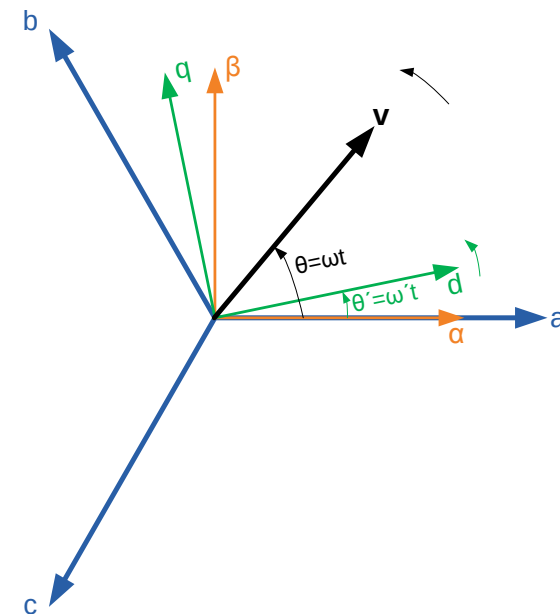
$$\begin{bmatrix} v_\alpha \\ v_\beta \end{bmatrix} = \begin{bmatrix} T_{\alpha\beta} \end{bmatrix} \begin{bmatrix} v_a \\ v_b \\ v_c \end{bmatrix} \quad \text{where } \theta' = 0$$

$$\begin{bmatrix} T_{\alpha\beta} \end{bmatrix} = \frac{2}{3} \begin{bmatrix} \cos(\theta') & \cos(\theta' - 120^\circ) & \cos(\theta' + 120^\circ) \\ -\sin(\theta') & -\sin(\theta' - 120^\circ) & -\sin(\theta' + 120^\circ) \end{bmatrix} = \frac{2}{3} \begin{bmatrix} 1 & -\frac{1}{2} & -\frac{1}{2} \\ 0 & \frac{\sqrt{3}}{2} & -\frac{\sqrt{3}}{2} \end{bmatrix}$$

αβ-frame → dq-frame

$$\begin{bmatrix} v_d \\ v_q \end{bmatrix} = \begin{bmatrix} T_{dq^{n-m}} \end{bmatrix} \begin{bmatrix} v_\alpha \\ v_\beta \end{bmatrix} \quad \text{where } n - m = 1$$

$$\begin{bmatrix} T_{dq^{n-m}} \end{bmatrix} = \begin{bmatrix} T_{dq^{m-n}} \end{bmatrix}^T = \begin{bmatrix} \cos(n-m)\omega' t & \sin(n-m)\omega' t \\ -\sin(n-m)\omega' t & \cos(n-m)\omega' t \end{bmatrix}$$



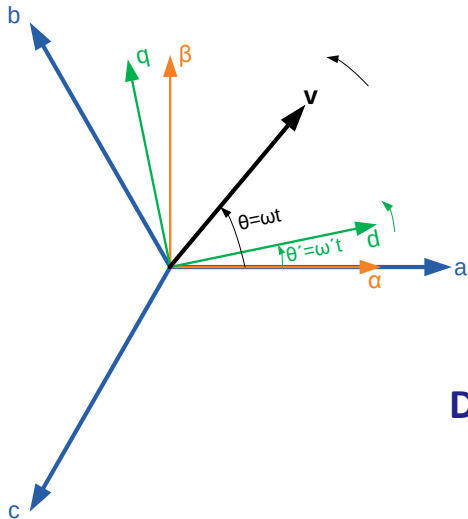


3. Grid Side Converter Controller

GSC Controller: Transformations – Synchronization – PQ controller – Current Controller – PWM

Park Transformations

- a-b-c natural frame
- $\alpha\beta$ -frame: stationary reference frame
- **dq-frame**: rotating with the synchronous rotational speed

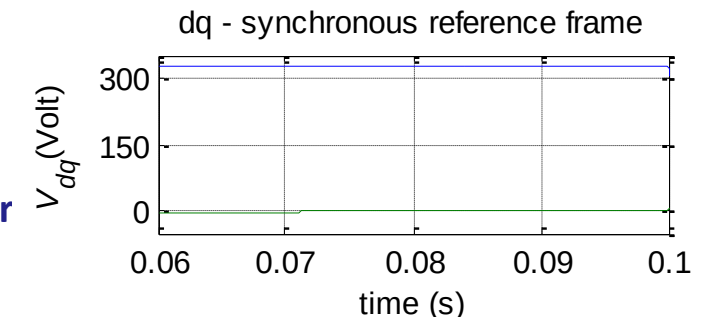
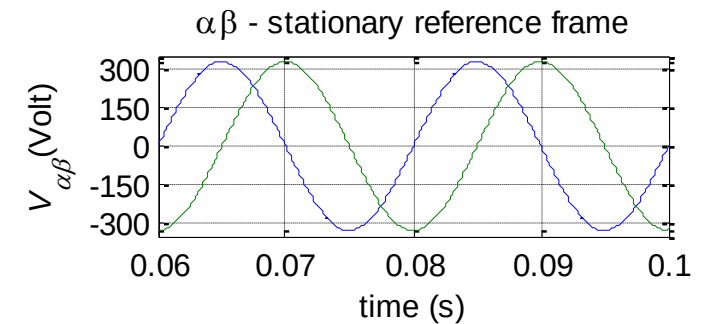
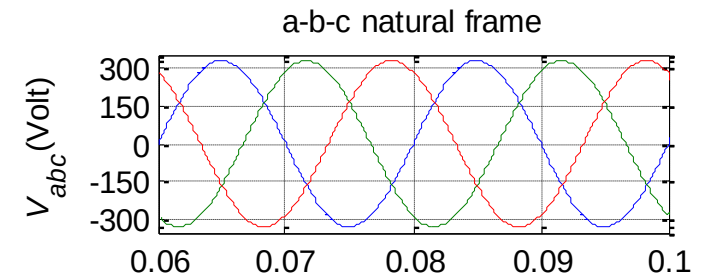


***dq-transformation enables AC voltages and currents to be transformed to DC quantities**



Easy to be controller (use of PI controller)

Decoupled control of active and reactive power

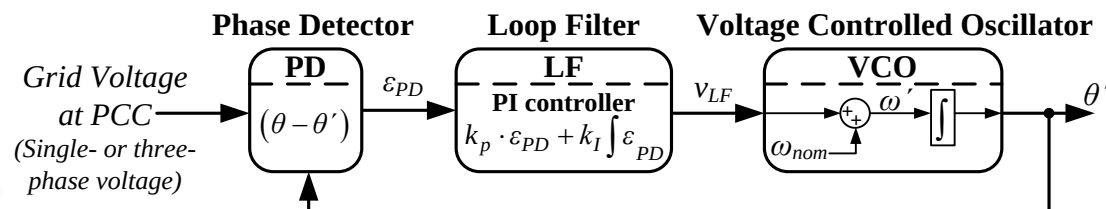


3. Grid Side Converter Controller

GSC Controller: Transformations – **Synchronization** – PQ controller – Current Controller – PWM

Synchronization Method

- The **most crucial unit** of the GSC controller
- Usually a phase locked loop algorithm (**PLL**)
- **Aim:** Dynamic and accurate **estimation of the voltage phase angle** (positive sequence phase angle of the fundamental voltage component)



• Important aspect:

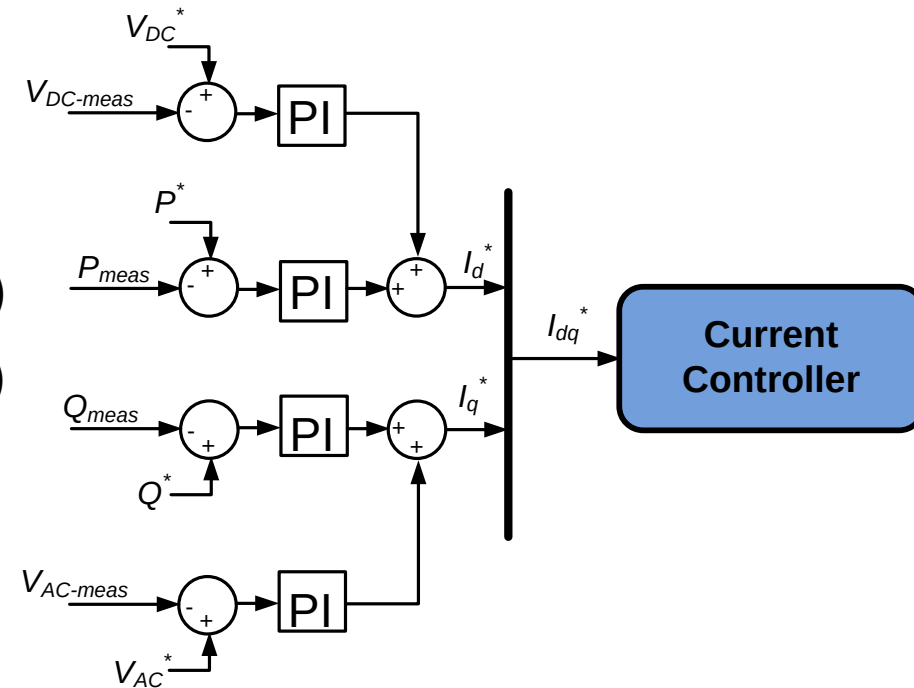
- Dynamic performance
- Accuracy against unbalanced voltage conditions (enable FRT operation under grid faults)
- Immunity against voltage harmonic distortion (enhanced the power quality of RES)
- Accuracy against frequency change
- Accuracy against phase jump
- Algorithms complexity (1/(Computational Efficiency))

3. Grid Side Converter Controller

GSC Controller: Transformations – Synchronization – **PQ controller** – Current Controller – PWM

PQ controller

- Control the active and reactive power injection
- Ensures that all the produced energy by the RES will be injected to the grid (Control of DC-link voltage)
- Ensures a voltage support to the grid (reactive power)
- Enhanced with FRT capabilities (meet grid codes)
- Limits the GSC currents within the converter ratings

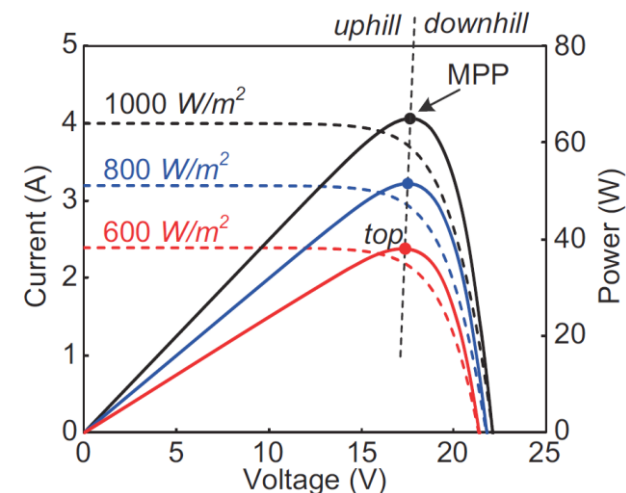


3. Grid Side Converter Controller

GSC Controller: Transformations – Synchronization – **PQ controller** – Current Controller – PWM

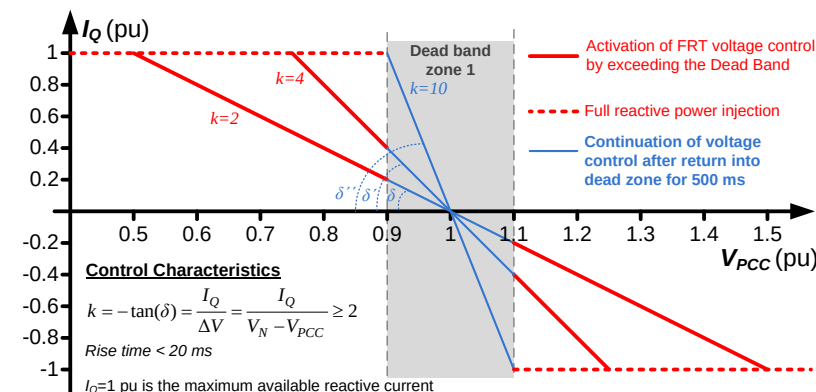
Maximum Power Point Tracking (PV only)

- Regulate the operating voltage of the PV module in order to maximize its power generation under variable sun irradiance conditions



Q-profile

- Inject or absorbs reactive power to support grid voltage (according to grid regulations)
- Voltage support under low-voltage grid faults



3. Grid Side Converter Controller

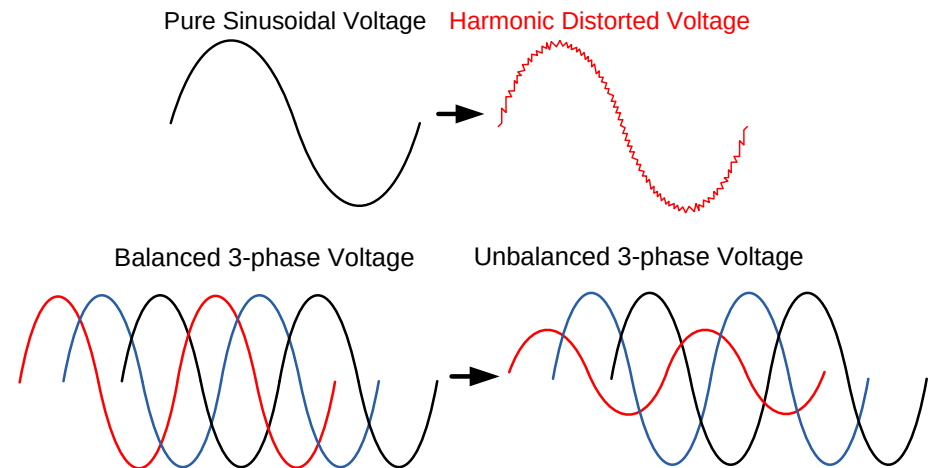
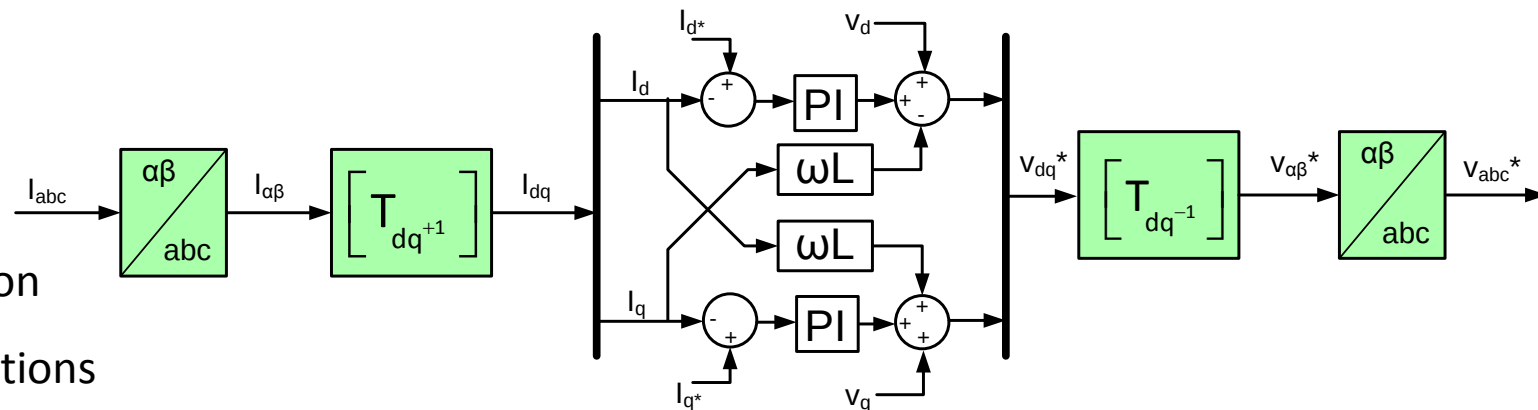
GSC Controller: Transformations – Synchronization – PQ controller – **Current Controller** – PWM

Current Controller

- Accurate current injection
- Immunity against harmonic distortion
- Immunity against unbalanced conditions

Conventional Current Controller

- Simplicity
- Design in dq-frame
- Problem under:
 - harmonic distortion
 - unbalanced conditions

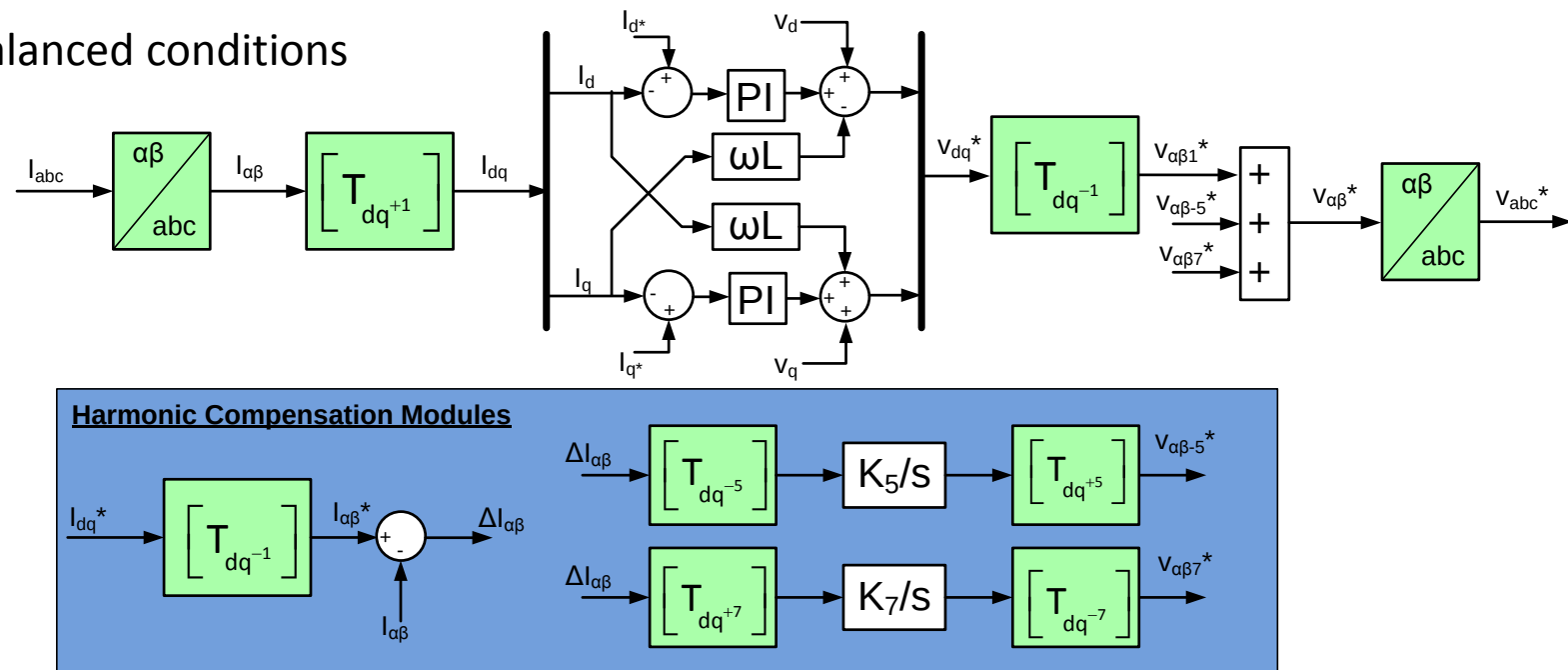


3. Grid Side Converter Controller

GSC Controller: Transformations – Synchronization – PQ controller – **Current Controller** – PWM

Current Controller enhanced with harmonic compensation modules

- Modularized
- Design in multiple dqⁿ-frames
- Accurate under harmonic distortion (Harmonic Compensation Modules)
- Problem under unbalanced conditions

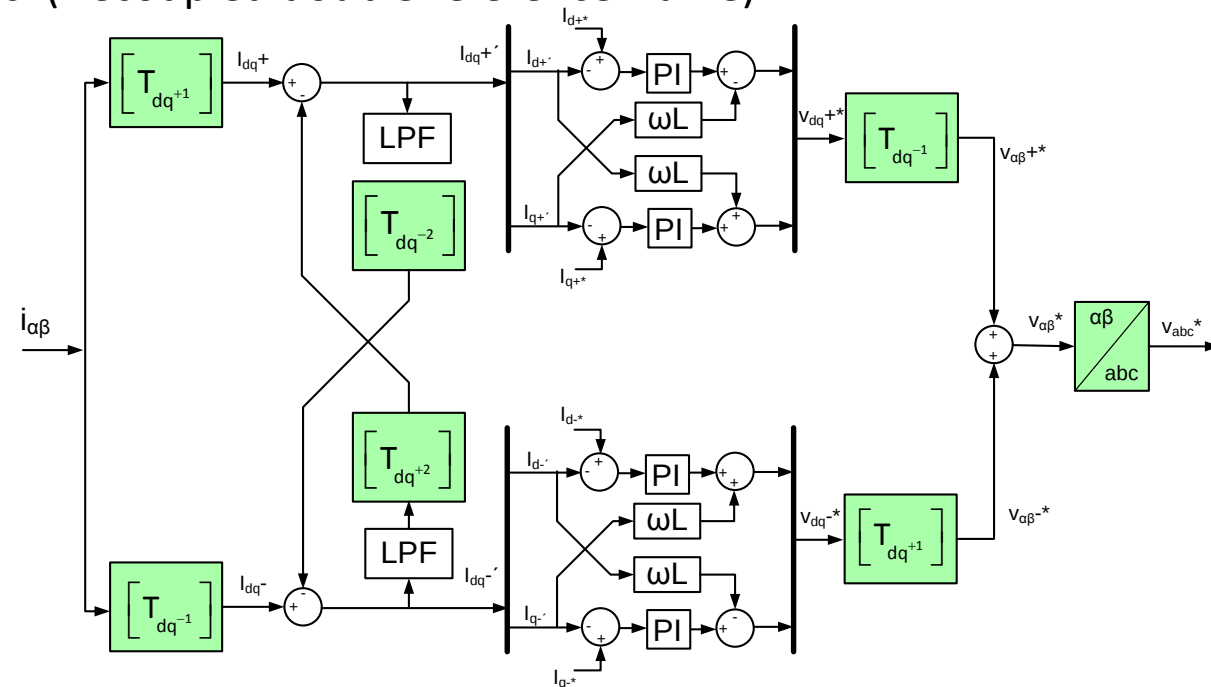


3. Grid Side Converter Controller

GSC Controller: Transformations – Synchronization – PQ controller – **Current Controller** – PWM

Decoupled double synchronous reference frame current controller

- Complex
- Design in positive and negative dqⁿ-frames
- Accurate under unbalanced conditions (Decoupled double reference frame)
- Problems under harmonic distortion



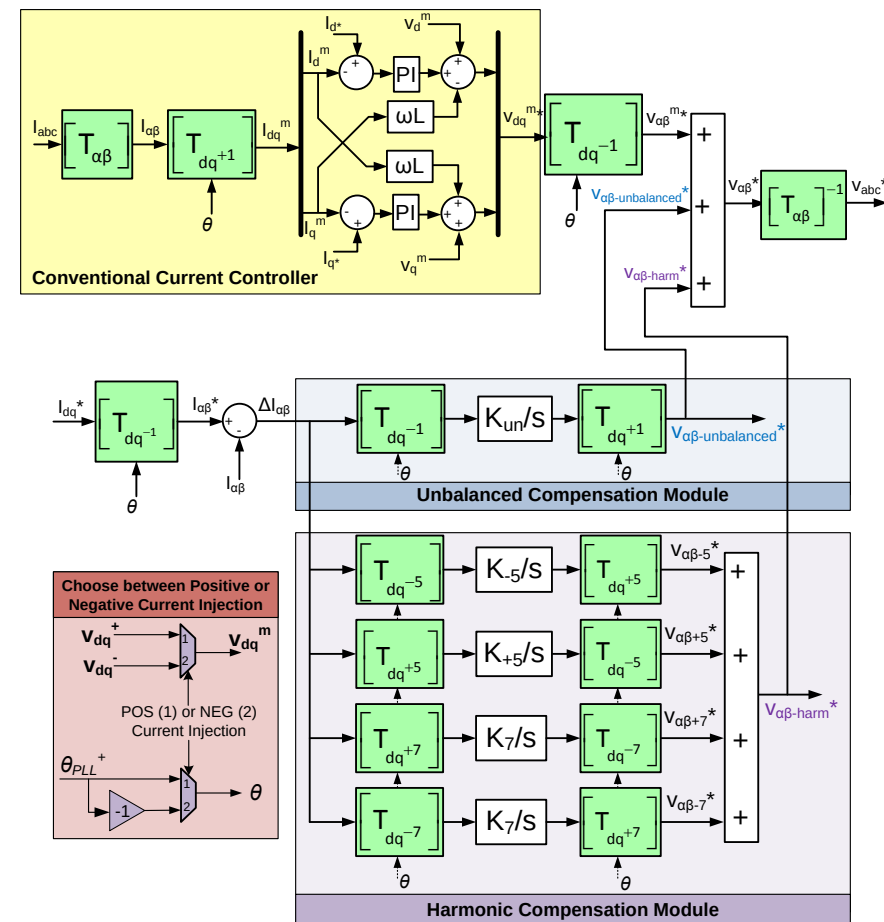


3. Grid Side Converter Controller

GSC Controller: Transformations – Synchronization – PQ controller – **Current Controller** – PWM

A proposed current controller enhanced with harmonic and unbalanced compensation module

- Simplicity and modularized
- Design in multiple dqⁿ-frames
- Accurate under unbalanced conditions (UCM)
- Accurate under harmonic distortion (HCM)
- Enable the accurate injection of positive or negative sequence current





3. Grid Side Converter Controller

GSC Controller: Transformations – Synchronization – PQ controller – **Current Controller** – PWM

Experimental results for the proposed Current Controller

Conditions:

Grid Fault: Type C ($d=65\%$) at 1 sec

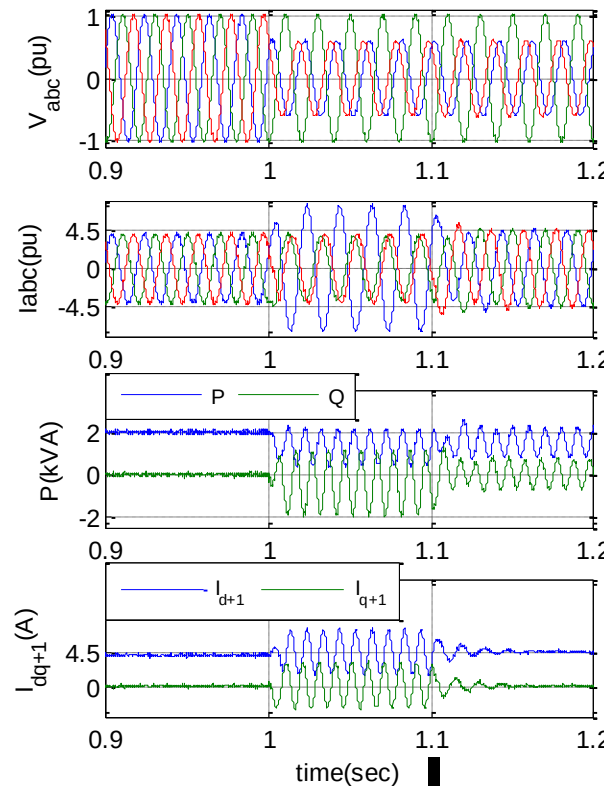
Harmonics: $|V_5|=2\%$ & $|V_7|=1\%$

FRT operation: $I_{\text{peak}} < 4.5\text{A}$ & $Q/P=3$

Proposed Current Controller

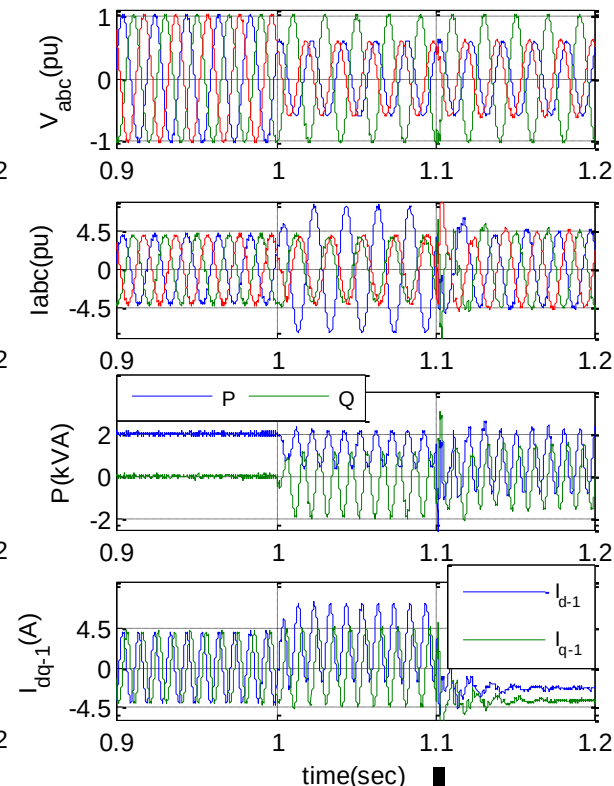
- HCM** is always activated
- UCM** is activated at 1.1 sec
- The UCM enables the **pure (+ or -) sinusoidal current injection** under FRT
- Fast and accurate current injection**
- Settling time < 10ms**
- High quality current injection** ($\text{THD}_i < 3\%$)

Full positive current injection



UCM is activated

Full negative current injection



UCM is activated

3. Grid Side Converter Controller

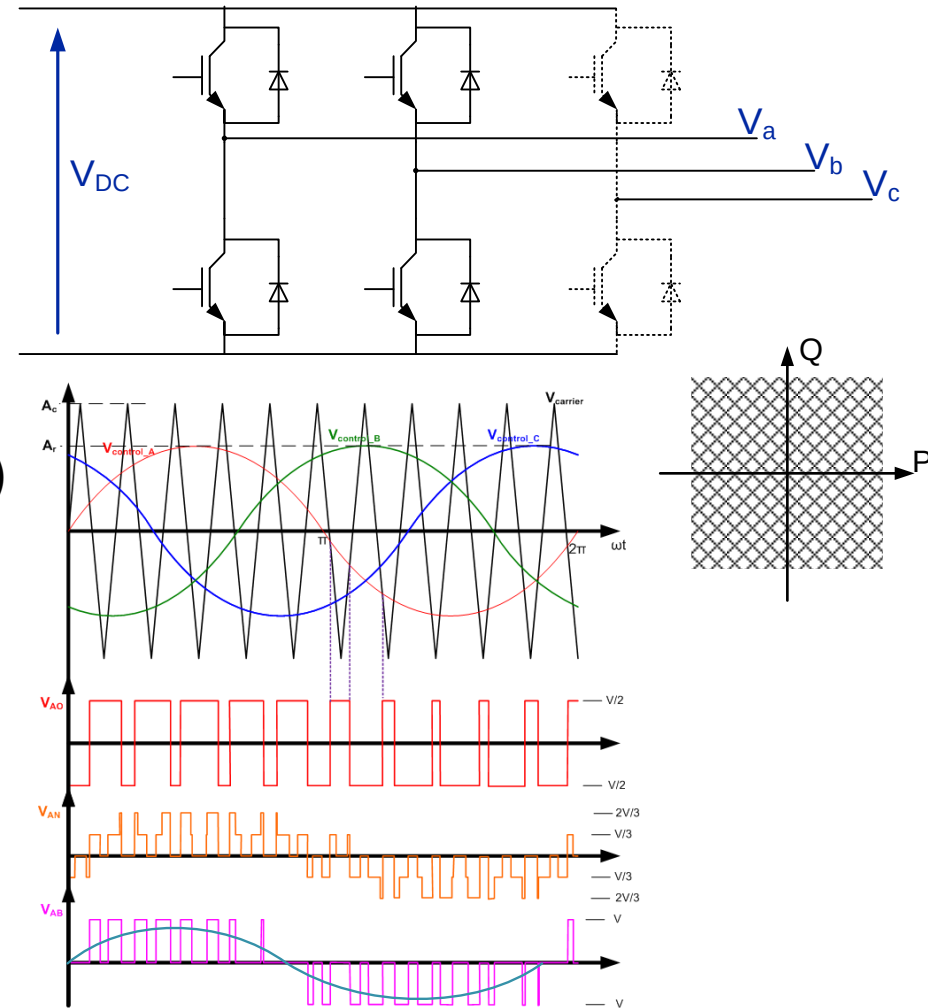
GSC Controller: Transformations – Synchronization – PQ controller – Current Controller – **PWM**

Grid Side power electronic Converter

- Single-phase: H-bridge topology, 4 IGBT or MOSFETs
- Three-phase: 2 level bidirectional Voltage Source Converter , 6 MOSFETs or IGBTs

Sinusoidal Pulse Width Modulation (SPWM)

- Control GSC through high frequency pulses (2-20kHz)
- A reference signal ($V_{control}$) is compared with a high frequency triangle signal (carrier)
- GSC produce a high frequency switching voltage
- LC or LCL filter eliminates high-order harmonics
- Amplitude Modulation $\rightarrow m_a = A_{control_A} / A_{carrier}$
- GSC output voltage $\rightarrow V_{L-L_RMS} = m_a \frac{\sqrt{3}V_{dc}}{2\sqrt{2}} = 0.612 \cdot m_a \cdot V_{dc}$

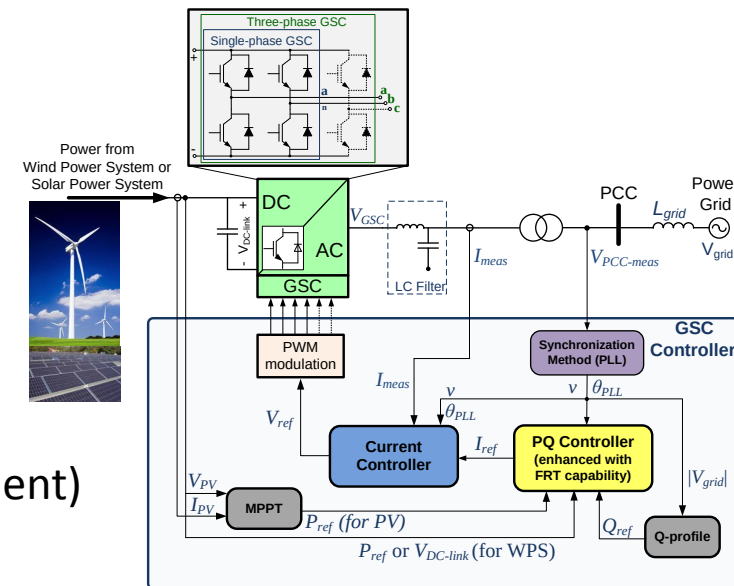


4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – Proposed methods

Synchronization Method / Phase-Locked Loop (PLL) algorithm

- The **most crucial unit** of the GSC controller
(The response of the PLL can affect the response of:
current controller, PQ controller and the entire GSC and RES)
- **Aim:** Dynamic and accurate estimation of the phase angle
(of the positive sequence of the fundamental voltage component)
- **Important aspect:**



- Dynamic performance
- Algorithms complexity (1/(Computational Efficiency))
- Accuracy against frequency change
- Accuracy against phase jump
- Accuracy against unbalanced voltage conditions (enable FRT operation under grid faults)
- Immunity against voltage harmonic distortion (enhanced the power quality of RES)

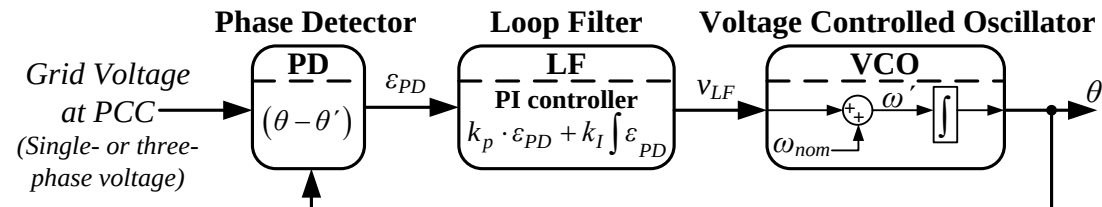
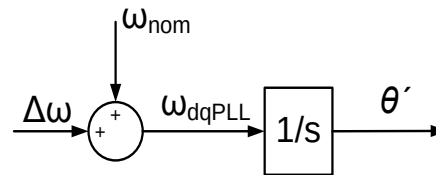
4. Synchronization Methods

Three-phase synchronization method: **Conventional methods** – State of the art – Proposed methods

A simplified PLL algorithm

- Phase Detector: $\theta - \theta'$
- Loop filter: Proportional-Integral (PI) controller
- Voltage Controller Oscillator

$$PI \text{ controller} \rightarrow Output_{PI} = error \left(k_P + k_I \frac{1}{s} \right)$$



Structure of a simplified PLL algorithm

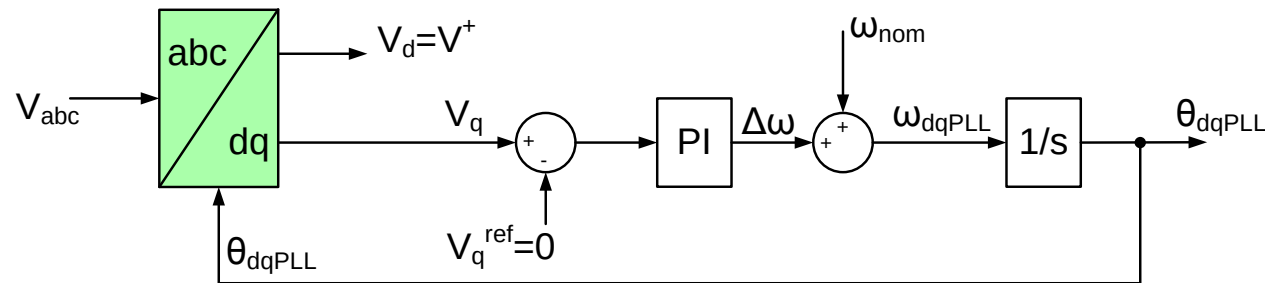
4. Synchronization Methods

Three-phase synchronization method: **Conventional methods** – State of the art – Proposed methods

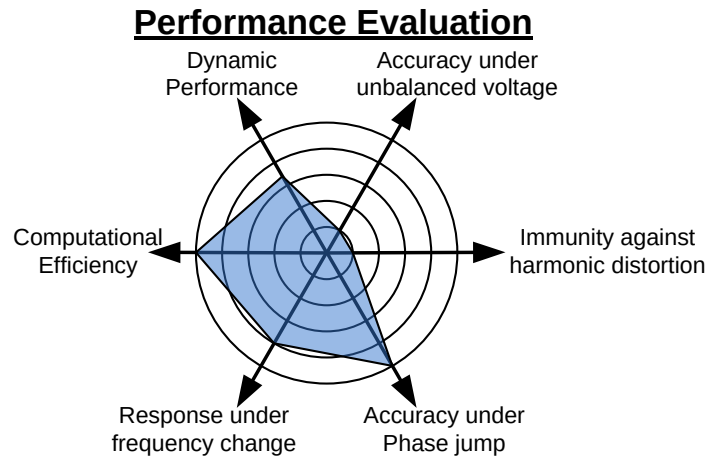
dq-PLL

- Design in dq-frame

- Phase Detector: $\theta - \theta'$ $\rightarrow V_{dq^{+1}}^{*+1} \approx V_{dq^{+1}}^{+1} = \begin{bmatrix} T_{dq^{+1}}(\theta') \end{bmatrix} \cdot v_{\alpha\beta}^{+1} = \begin{bmatrix} \cos \theta' & \sin \theta' \\ -\sin \theta' & \cos \theta' \end{bmatrix} \begin{bmatrix} V^1 \cdot \cos \theta \\ V^1 \cdot \cos(\theta - T/4) \end{bmatrix} = \begin{bmatrix} V^1 \cdot \cos(\theta - \theta') \\ V^1 \cdot \sin(\theta - \theta') \end{bmatrix} \approx \begin{bmatrix} V^1 \\ V^1 \Delta \theta \end{bmatrix}$



Structure of the dq-PLL

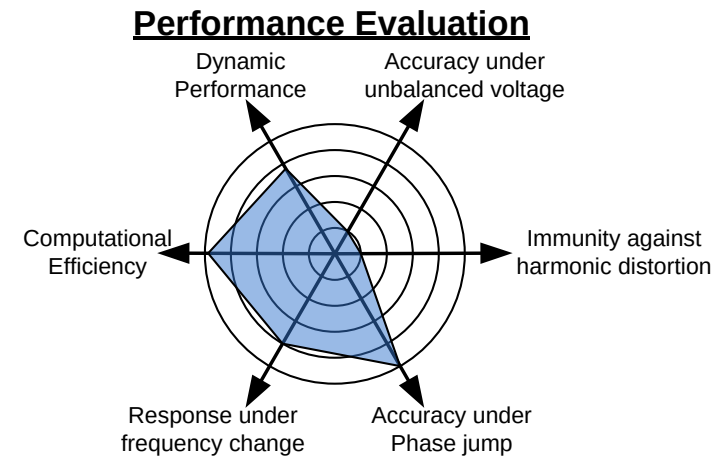
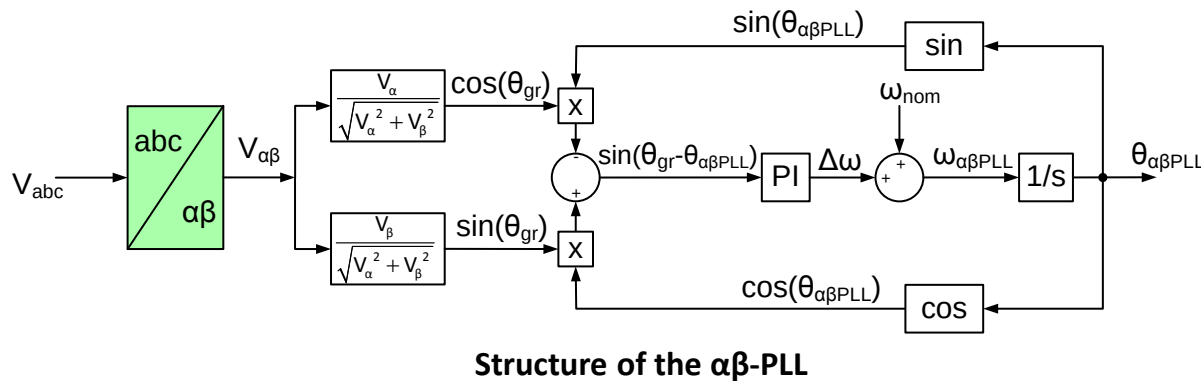


4. Synchronization Methods

Three-phase synchronization method: **Conventional methods** – State of the art – Proposed methods

$\alpha\beta$ -PLL

- Design in $\alpha\beta$ -frame
- Phase Detector: $\theta - \theta'$ $\rightarrow$ $\Delta\theta = \theta_{gr} - \theta_{ab-PLL} \approx \sin(\theta_{gr} - \theta_{\alpha\beta-PLL}) = \sin(\theta_{gr})\cos(\theta_{\alpha\beta-PLL}) - \sin(\theta_{\alpha\beta-PLL})\cos(\theta_{gr})$

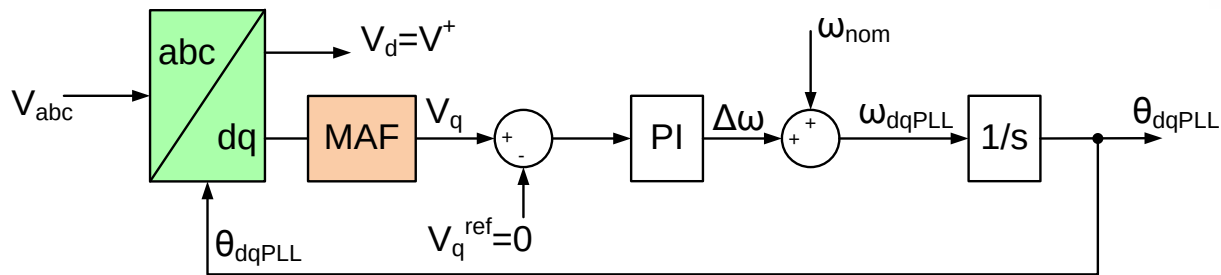


4. Synchronization Methods

Three-phase synchronization method: Conventional methods – **State of the art** – Proposed methods

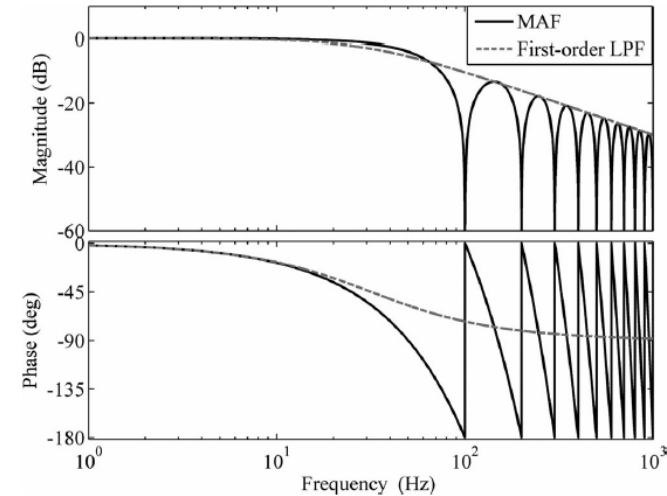
MAF-PLL (Moving Average Filter PLL)

- Inherits the structure of the dq-PLL
- Insert a MAF to minimize the effect of unbalanced and harmonic distorted voltage

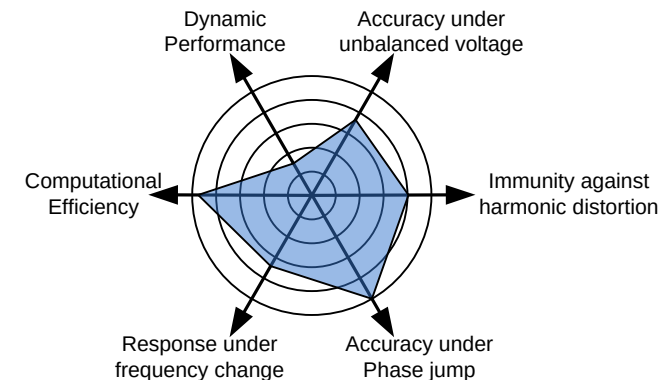


Structure of the MAF-PLL

Bode Diagram of MAF



Performance Evaluation

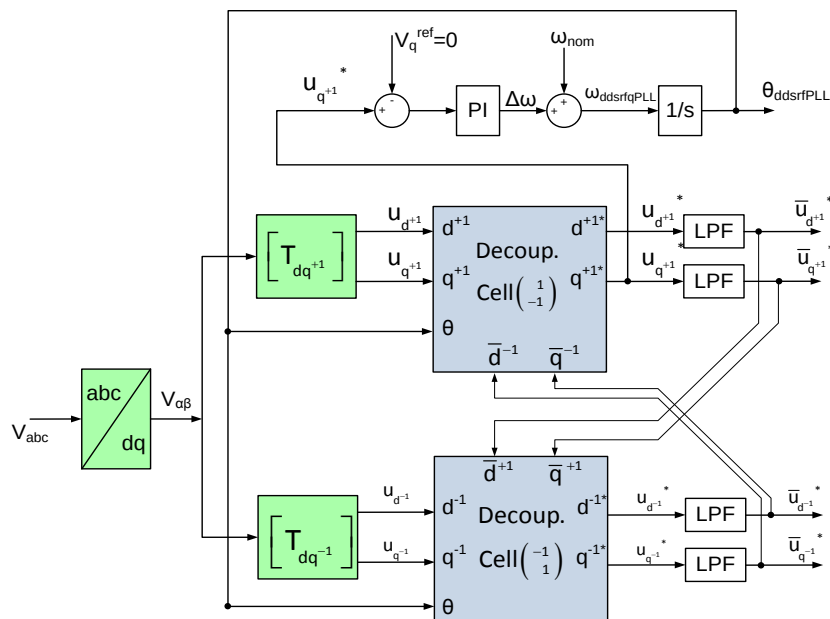


4. Synchronization Methods

Three-phase synchronization method: Conventional methods – **State of the art** – Proposed methods

ddsrf-PLL (Decoupled Double Synchronous Reference Frame PLL)

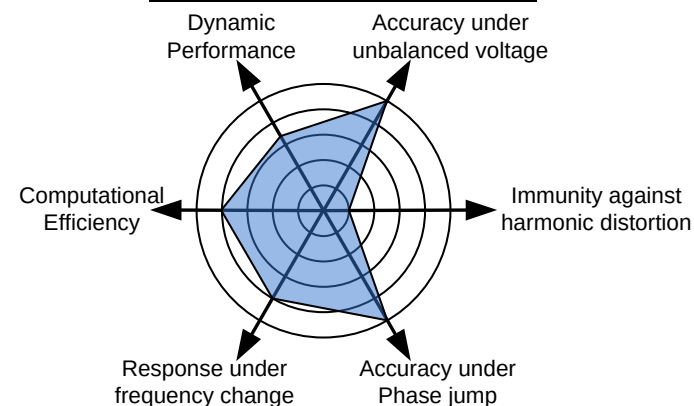
- Design in two dq-frame: dq^{+1} -frame & dq^{-1} -frame
- Phase Detector: $\theta - \theta'$ → The same one with the dq-PLL
- Decoupling Cell to decouple the effect of negative sequence (unbalanced voltage)



Structure of the dsrf-PLL

$$u_{dq^{+1}} = \begin{bmatrix} u_{d^{+1}} \\ u_{q^{+1}} \end{bmatrix} = \begin{bmatrix} T_{dq^{+1}} \end{bmatrix} \cdot u_{ab} = \underbrace{V^{+1}}_{dc-term} \begin{bmatrix} 1 \\ 0 \end{bmatrix} + V^{-1} \begin{bmatrix} \cos(-2\omega t) \\ \sin(-2\omega t) \end{bmatrix}$$

Performance Evaluation

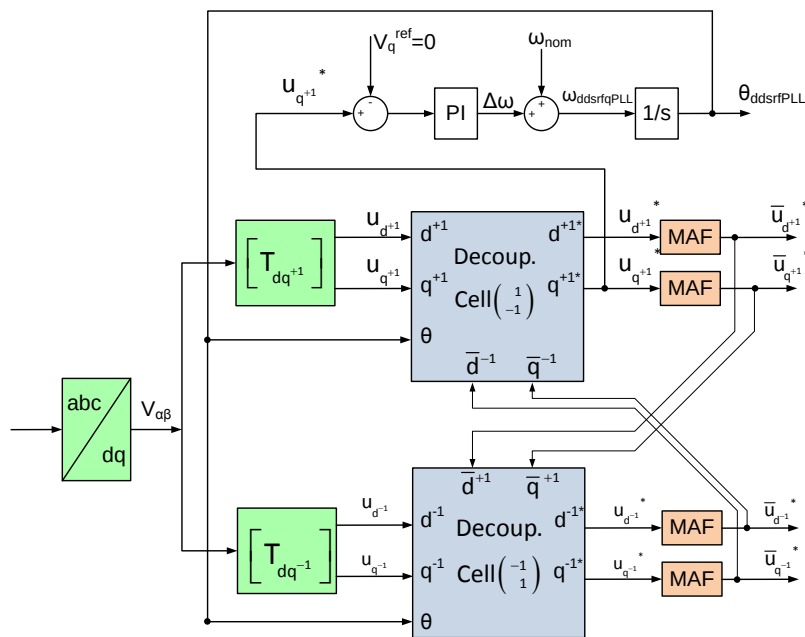


4. Synchronization Methods

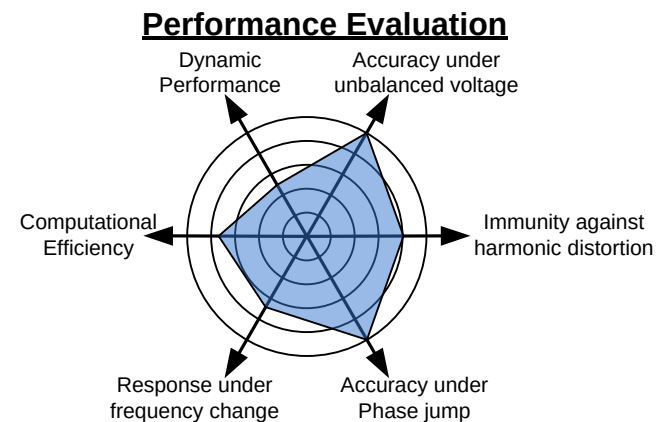
Three-phase synchronization method: Conventional methods – **State of the art** – Proposed methods

Modified MRF/MAF-PLL (Modified Multi Reference Frame / Moving Average Filter PLL)

- Inherits the structure of the ddsrf-PLL
- The Decoupling Cells eliminate the effect of unbalanced voltage
- Use of four MAF instead of LPFs to minimize the effect of harmonic distorted voltage



Structure of the MRF/MAF-PLL

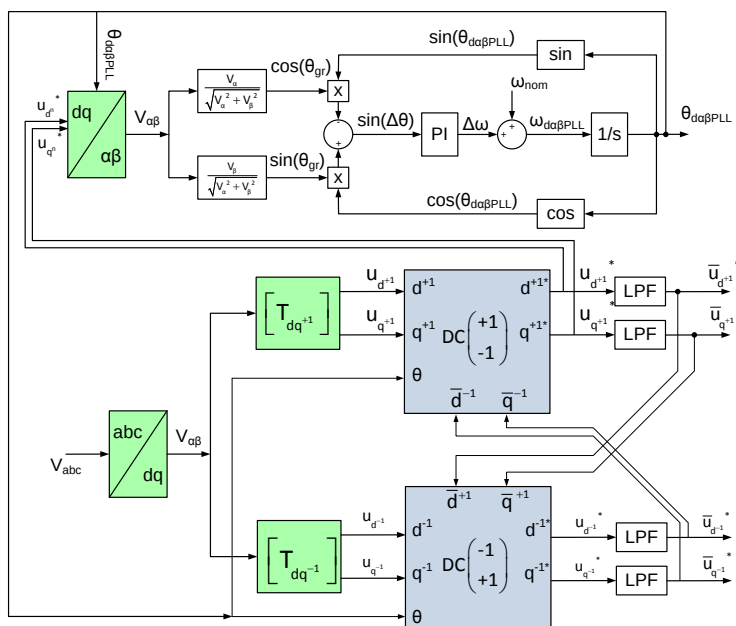


4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

$\alpha\beta$ -PLL (Decoupled in $\alpha\beta$ stationary reference Frame PLL)

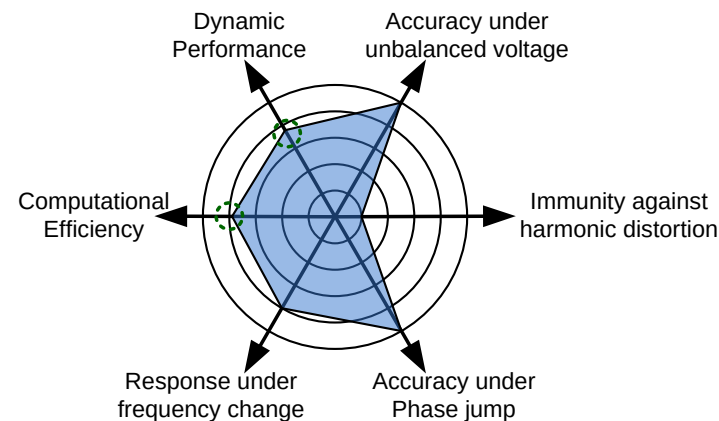
- Design in two dq-frame: dq^{+1} -frame & dq^{-1} -frame
- Phase Detector: $\theta - \theta'$ → Inherits the phase detector method of $\alpha\beta$ -PLL
- Decoupling Cell to decoupled the effect of negative sequence voltage (Same as ddsrf-PLL)



Structure of the $\alpha\beta$ -PLL

$$u_{dq^{+1}} = \begin{bmatrix} u_{d^{+1}} \\ u_{q^{+1}} \end{bmatrix} = \begin{bmatrix} T_{dq^{+1}} \end{bmatrix} \cdot u_{ab} = \underbrace{V^{+1} \begin{bmatrix} 1 \\ 0 \end{bmatrix}}_{dc-term} + V^{-1} \begin{bmatrix} \cos(-2\omega t) \\ \sin(-2\omega t) \end{bmatrix}$$

Performance Evaluation



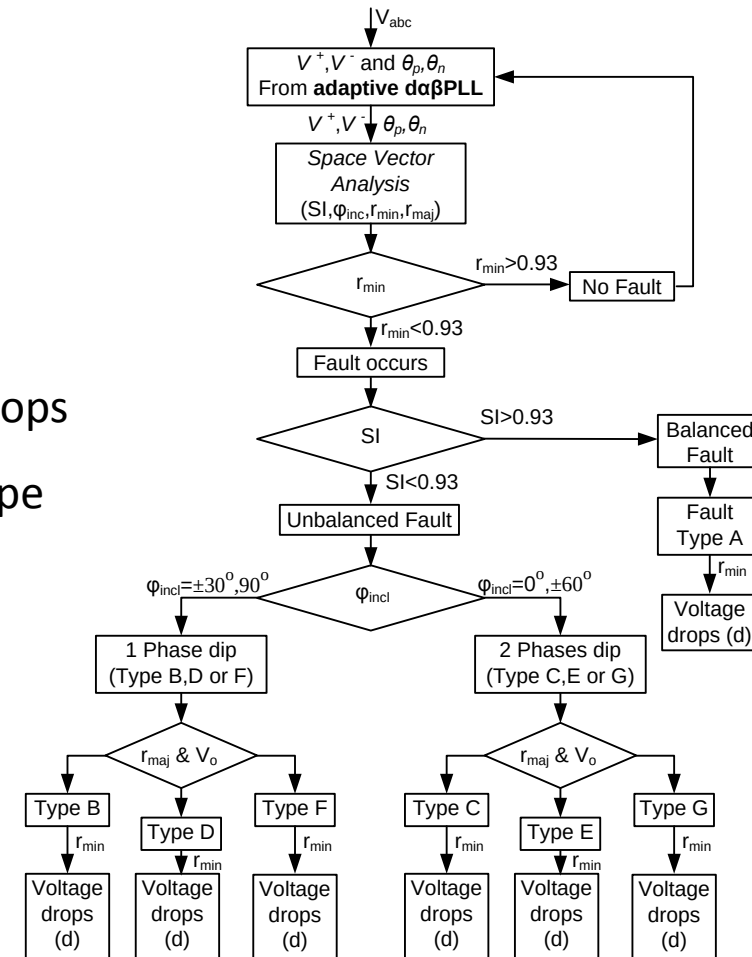
4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

Adaptive $\alpha\beta$ -PLL

- Inherits the structure of the $\alpha\beta$ -PLL
- Proposed tuning mechanism $\rightarrow$ adaptively change the tuning parameters (k_p , T_I) according to the type of the grid fault in order to accelerate the performance of the PLL
- A fault classification unit detect the type and the voltage droops
- The tuning parameters are adjusted according to the fault type based on a look-up table

Online Fault Classification Unit

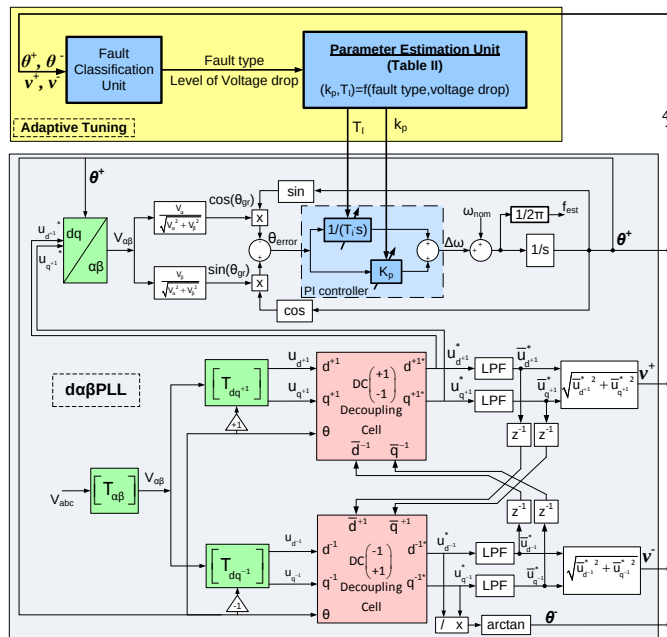


4. Synchronization Methods

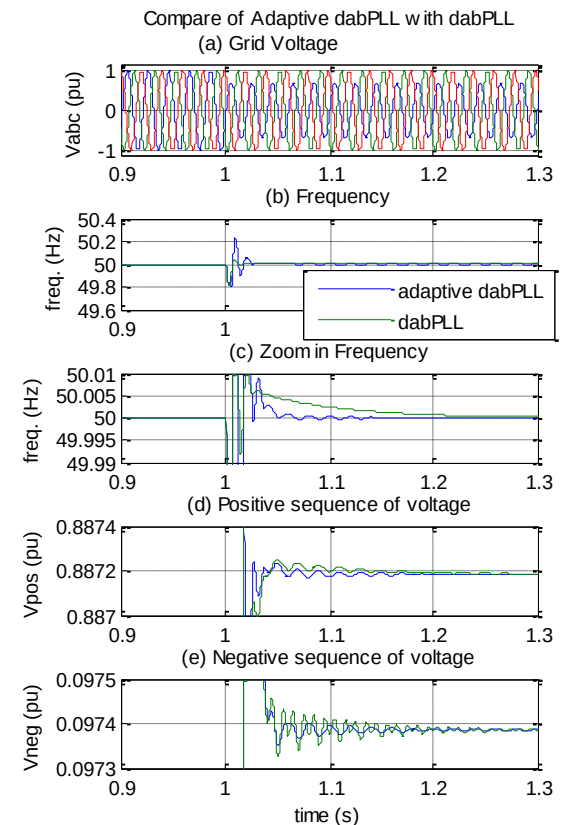
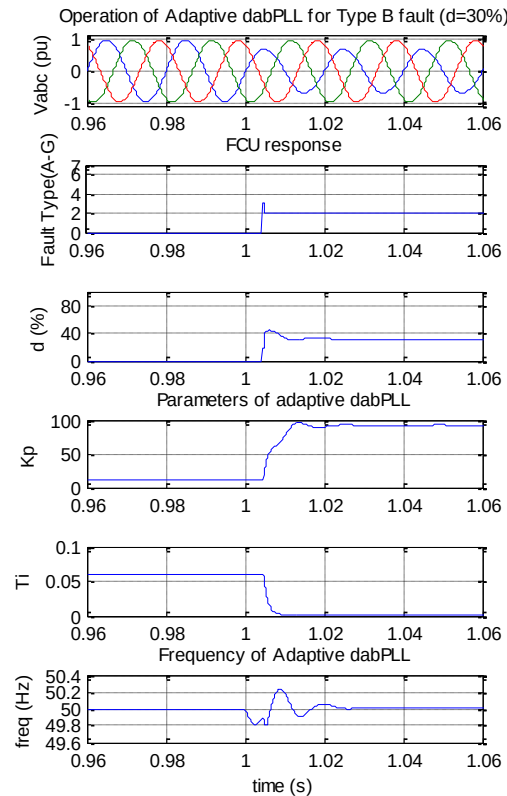
Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

Adaptive $\alpha\beta$ -PLL

- Proposed tuning mechanism $\rightarrow$ adaptively change the tuning parameters (k_p , T_i) of the $\alpha\beta$ -PLL according to the FCU to accelerate the performance of the PLL



Structure of the adaptive $\alpha\beta$ -PLL

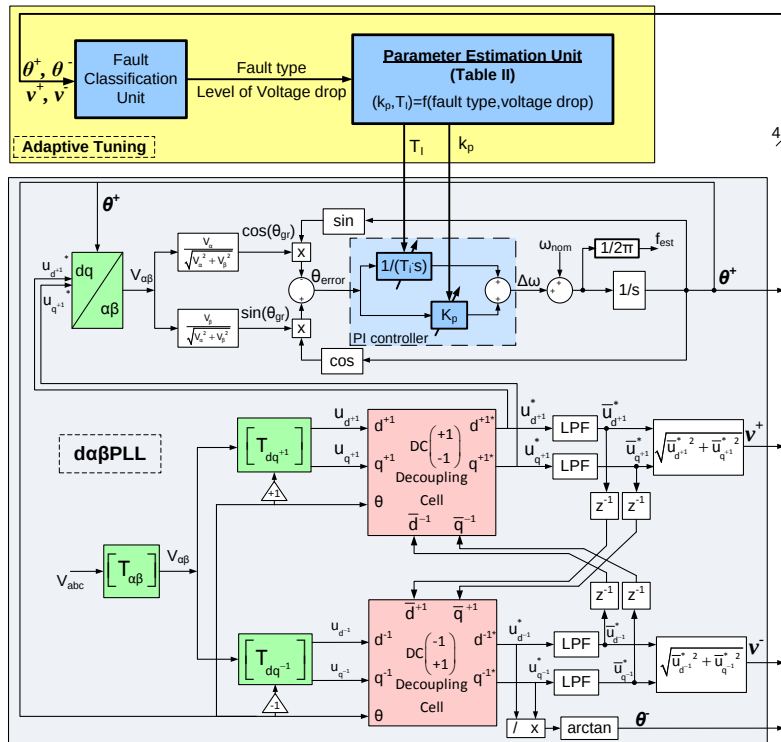


4. Synchronization Methods

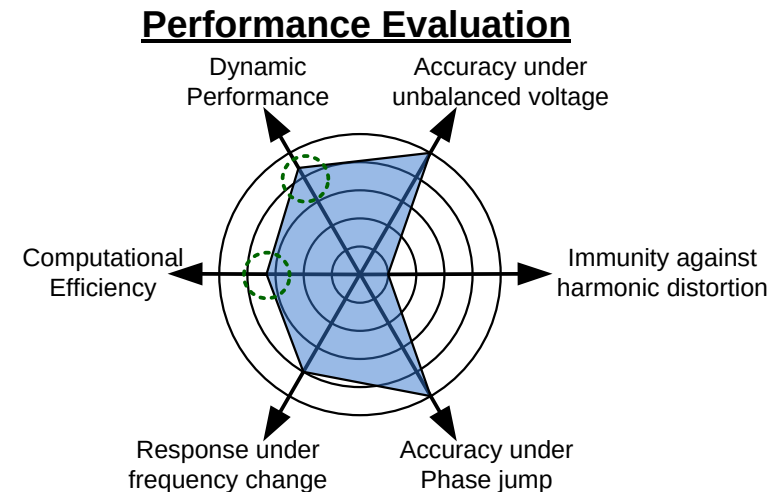
Three-phase synchronization method: Conventional methods – State of the art – Proposed methods

Adaptive $d\alpha\beta$ -PLL

- Proposed tuning mechanism $\rightarrow$ adaptively change the tuning parameters (k_p , T_I) of the $\alpha\beta$ -PLL according to the FCU to accelerate the performance of the PLL



Structure of the adaptive $d\alpha\beta$ -PLL

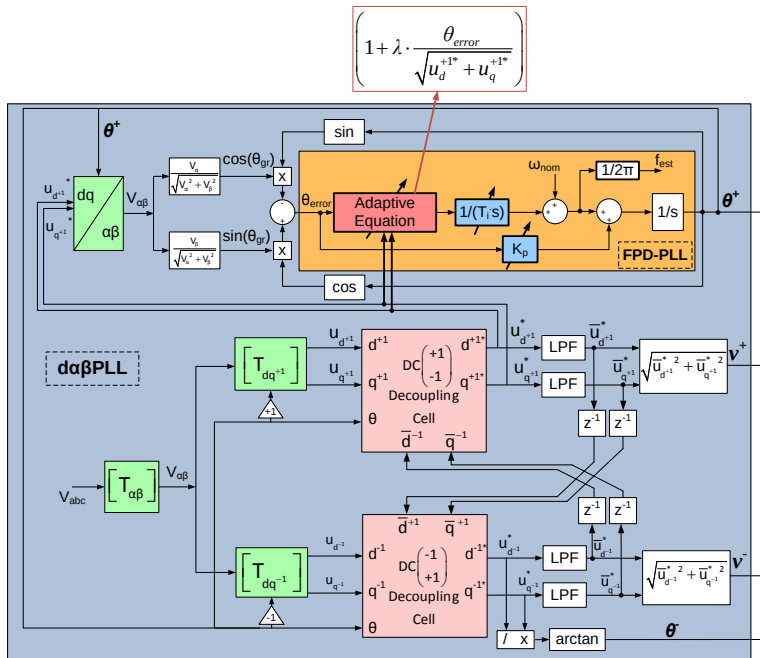


4. Synchronization Methods

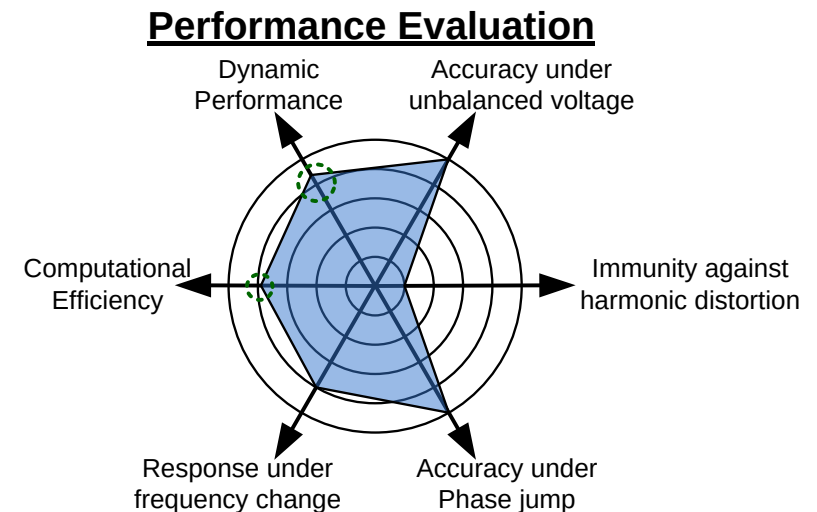
Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

FPD- $\alpha\beta$ -PLL (Frequency phase decoupling PLL)

- Inherits the structure of $\alpha\beta$ -PLL
- The tuning parameters (T_f) is adjusted according to the disturbance faced by the PLL
- Reconstruction of the PI controller in order to decouple the frequency and the phase estimation



Structure of the FPD- $\alpha\beta$ -PLL

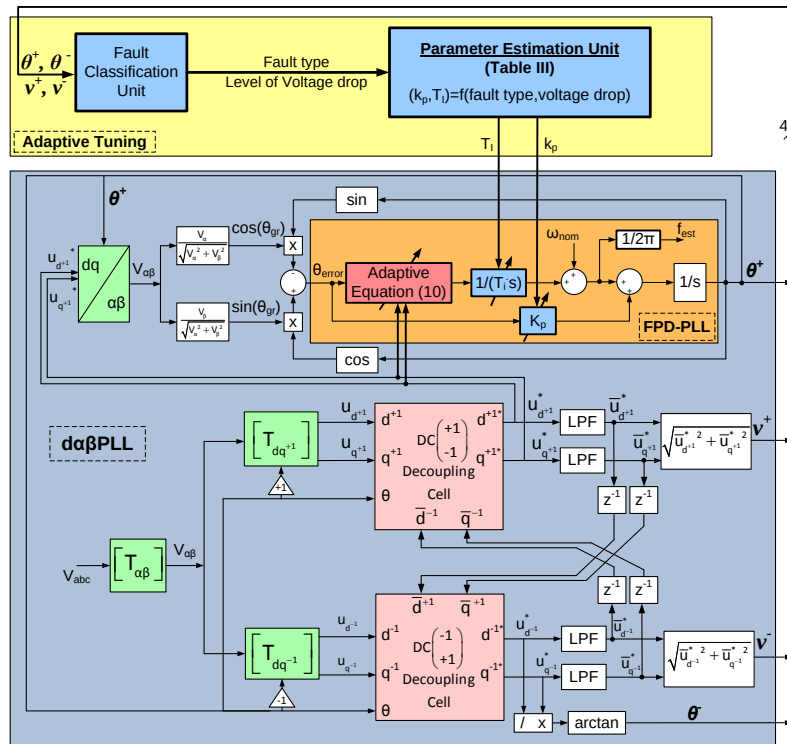


4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

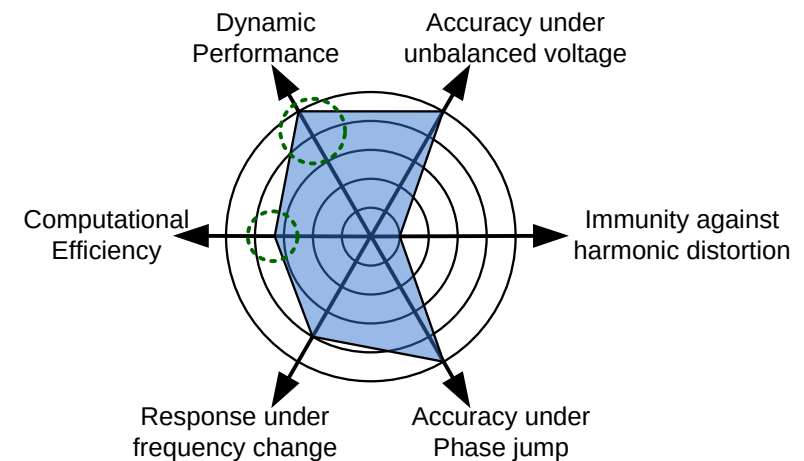
Adaptive FPD- $\alpha\beta$ -PLL (Adaptive Frequency phase decoupling PLL)

- Inherits the structure of $\alpha\beta$ -PLL
- Inherits the tuning mechanism of adaptive $\alpha\beta$ -PLL & the tuning mechanism of FPD- $\alpha\beta$ -PLL



Structure of the Adaptive FPD- $\alpha\beta$ -PLL

Performance Evaluation



4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

MSHDC-PLL (Multi-Sequence/Harmonic Decoupling Cell PLL)

- Extend the decoupling cell of the $d\alpha\beta$ -PLL to consider unbalanced and harmonic conditions
- Voltage analysis under unbalanced and harmonic distorted (5th) conditions

$$u_{dq^{+1}} = \begin{bmatrix} u_{d^{+1}} \\ u_{q^{+1}} \end{bmatrix} = \begin{bmatrix} T_{dq^{+1}} \end{bmatrix} \cdot u_{ab} = \underbrace{V^{+1} \begin{bmatrix} 1 \\ 0 \end{bmatrix}}_{dc-term} + \underbrace{V^{-1} \begin{bmatrix} \cos(-2\omega t) \\ \sin(-2\omega t) \end{bmatrix}}_{-2\omega \text{ oscillation}} + \underbrace{V^{+5} \begin{bmatrix} \cos((+1-5)\omega t) \\ \sin((+1-5)\omega t) \end{bmatrix}}_{-4\omega \text{ oscillation}} + \underbrace{V^{-5} \begin{bmatrix} \cos((+1+5)\omega t) \\ \sin((+1+5)\omega t) \end{bmatrix}}_{+6\omega \text{ oscillation}}$$

- Express the voltage in each reference frame

$$\begin{bmatrix} v_{dq^{+1}} \\ v_{dq^{-1}} \\ v_{dq^{+5}} \\ v_{dq^{-5}} \end{bmatrix} = \begin{bmatrix} T_{dq^{+1}} \\ T_{dq^{-1}} \\ T_{dq^{+5}} \\ T_{dq^{-5}} \end{bmatrix} v_{\alpha\beta} = \underbrace{\begin{bmatrix} V_{dq}^{+1} \\ V_{dq}^{-1} \\ V_{dq}^{+5} \\ V_{dq}^{-5} \end{bmatrix}}_{DC \text{ terms}} + \underbrace{\begin{bmatrix} [0] & T_{dq^{+1}-(+1)} & T_{dq^{+1}-(+5)} & T_{dq^{+1}-(-5)} \\ T_{dq^{-1}-(+1)} & [0] & T_{dq^{-1}-(+5)} & T_{dq^{-1}-(-5)} \\ T_{dq^{+5}-(+1)} & T_{dq^{+5}-(+5)} & [0] & T_{dq^{+5}-(-5)} \\ T_{dq^{-5}-(+1)} & T_{dq^{-5}-(+5)} & T_{dq^{-5}-(-5)} & [0] \end{bmatrix}}_{Oscillation \text{ Terms}} \begin{bmatrix} V_{dq}^{+1} \\ V_{dq}^{-1} \\ V_{dq}^{+5} \\ V_{dq}^{-5} \end{bmatrix}$$

- Include this equation on a cross feedback Multi-Sequence/Harmonic Decoupling Cell (MSHDC) in order to estimate **accurately** and **dynamically** the (dc-term) amplitude of each voltage harmonic

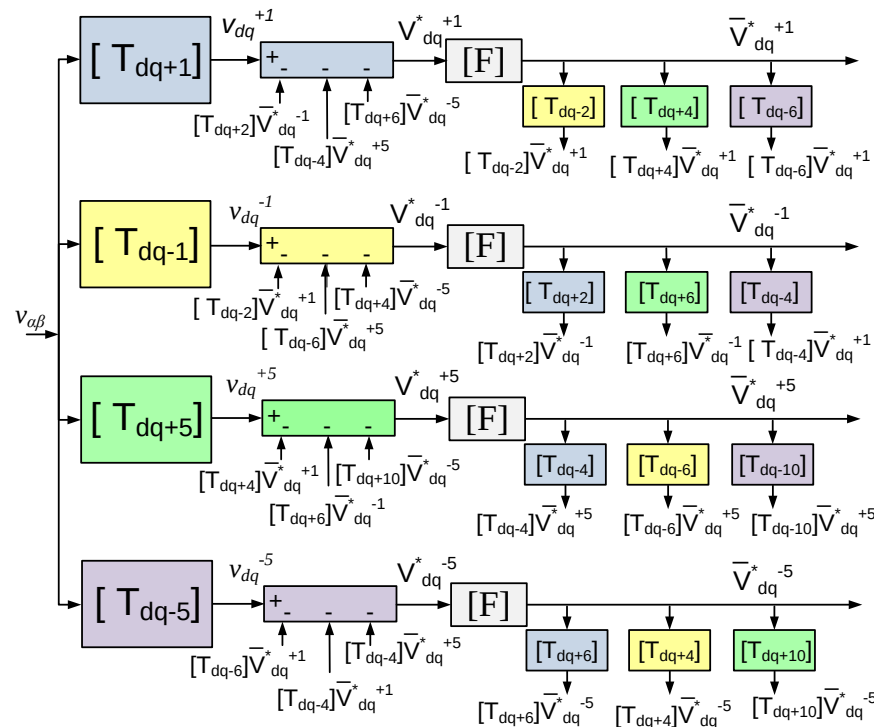
4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

MSHDC-PLL (Multi-Sequence/Harmonic Decoupling Cell PLL)

• Decoupling Network $\rightarrow$

$$\bar{V}_{dq}^{*(n)} = [F] \left\{ \begin{bmatrix} u_d^{(n)} \\ u_q^{(n)} \end{bmatrix} - \sum_{m \neq n} \left\{ T_{dq^{(n-m)}} \cdot \begin{bmatrix} \bar{V}_{dq}^{*(m)} \\ \bar{V}_{dq}^{*(m)} \end{bmatrix} \right\} \right\} = \bar{V}_{dq}^{*(n)} = [F] \left\{ T_{dq^{(n)}} \begin{bmatrix} v_a \\ v_\beta \end{bmatrix} - \sum_{m \neq n} \left\{ T_{dq^{(n-m)}} \begin{bmatrix} \bar{V}_{dq}^{*(m)} \\ \bar{V}_{dq}^{*(m)} \end{bmatrix} \right\} \right\}$$

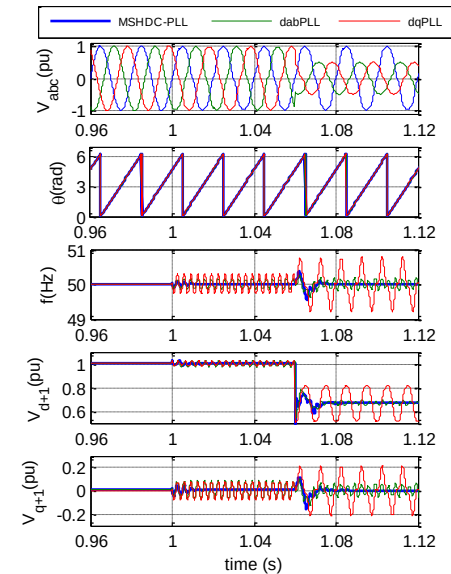
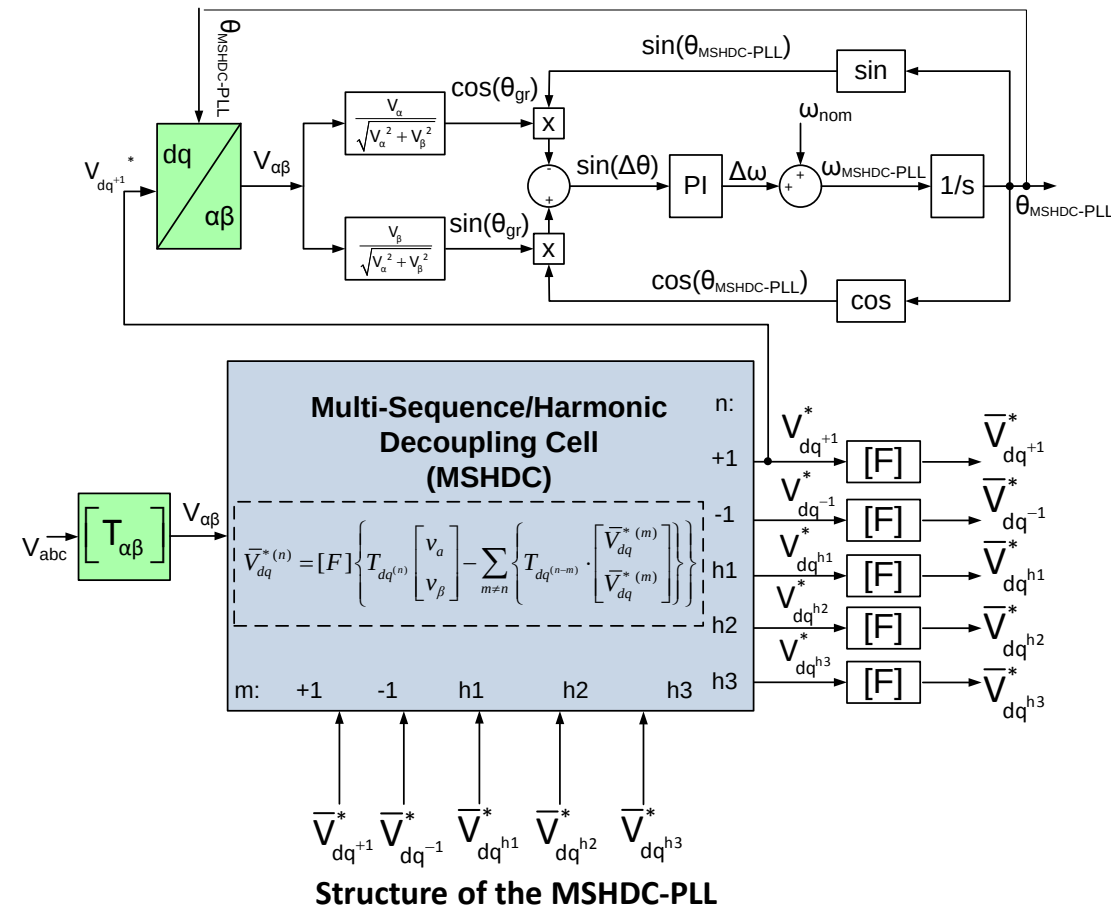


Structure of the MSHDC

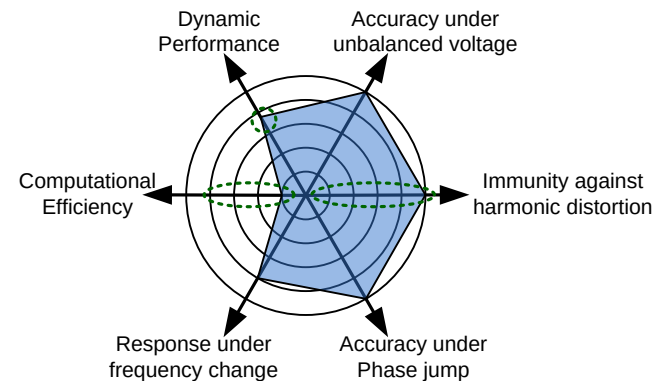
4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

MSHDC-PLL (Multi-Sequence/Harmonic Decoupling Cell PLL)



Performance Evaluation



4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

DNαβ-PLL (DN in αβ-frame PLL)

- MSHDC-PLL present very good accuracy under unbalanced and harmonic conditions
- MSHDC present high complexity (required processing time)

Re-design the MSHDC in αβ-frame (simplifies the complexity of the algorithm)

$$\bullet \text{MSHDC} \rightarrow \bar{V}_{dq}^{*(n)} = [F] \left\{ \begin{bmatrix} u_d^{(n)} \\ u_q^{(n)} \end{bmatrix} - \sum_{m \neq n} \left\{ T_{dq^{(n-m)}} \cdot \begin{bmatrix} \bar{V}_{dq}^{*(m)} \\ \bar{V}_{dq}^{*(m)} \end{bmatrix} \right\} \right\} = \bar{V}_{dq}^{*(n)} = [F] \left\{ T_{dq^{(n)}} \begin{bmatrix} v_a \\ v_\beta \end{bmatrix} - \sum_{m \neq n} \left\{ T_{dq^{(n-m)}} \begin{bmatrix} \bar{V}_{dq}^{*(m)} \\ \bar{V}_{dq}^{*(m)} \end{bmatrix} \right\} \right\}$$

$$\bullet \text{DN}\alpha\beta \rightarrow v_{dq_n}^{*n} = \begin{bmatrix} T_{dq^n} \end{bmatrix} v_{a\beta}^{*n} = \begin{bmatrix} T_{dq^n} \end{bmatrix} \left(v_{a\beta} - \sum_{m \neq n} \begin{bmatrix} T_{dq^{-n}} \end{bmatrix} [F(s)] \begin{bmatrix} T_{dq^n} \end{bmatrix} v_{a\beta}^{*m} \right)$$

- **DNαβ achieve equivalent performance with the MSHDC,**
however the computational efficiency of the method
is sufficiently enhanced (61% less required processing time)

COMPLEXITY COMPARISON BETWEEN THE DNab AND THE MSHDC

Decoupling Network	Complexity analysis in each control loop			Total
	$\left(v_{dq_n} - \sum_{m \neq n} v_{dq_n}^{*m} \right) \text{ or } \left(v_{a\beta} - \sum_{m \neq n} v_{a\beta}^{*m} \right)$	$[T_{dq^n}]$	$[F(s)]$	
MSHDC (N=10)	N	N ²	N	640 Multiplications 120 Additions 280 Subtractions
DNαβ (N=10)	N	2N	N	160 Multiplications 40 Additions 200 Subtractions

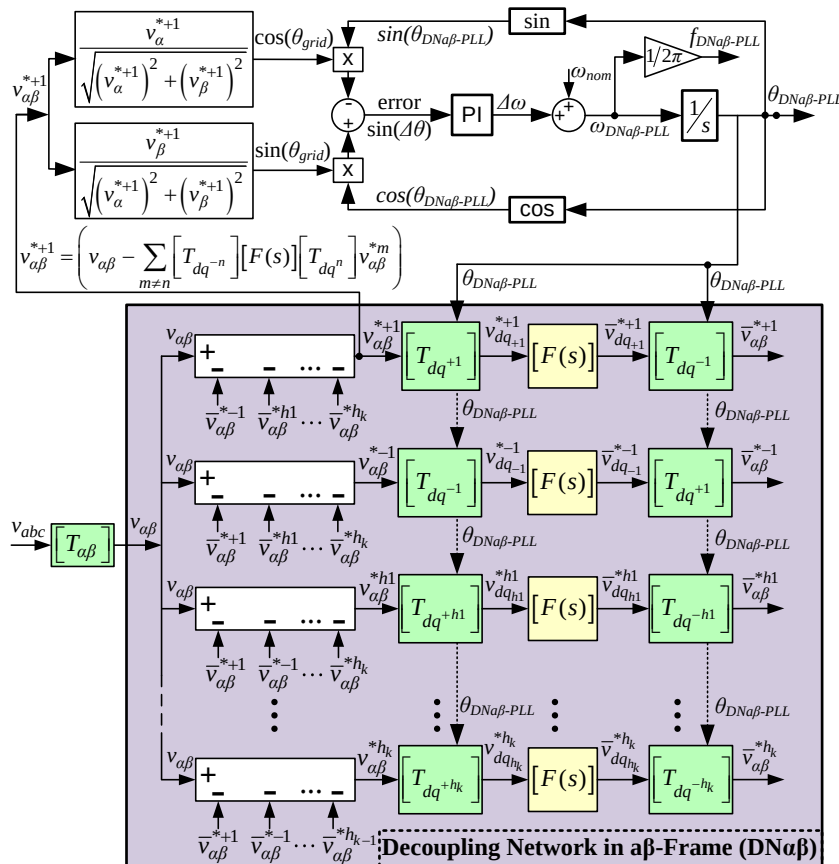
Notes:

- Each $(v_{a\beta} - \sum_{m \neq n} v_{a\beta}^{*m})$ or $(v_{dq_n} - \sum_{m \neq n} v_{dq_n}^{*m}) \rightarrow$ requires 2(N-1) Subtractions
- Each $[T_{dq}] \rightarrow$ requires 6 Multiplications + 1 Addition + 1 Subtractions
- Each $[F(s)] \rightarrow$ requires 4 Multiplications + 2 Additions

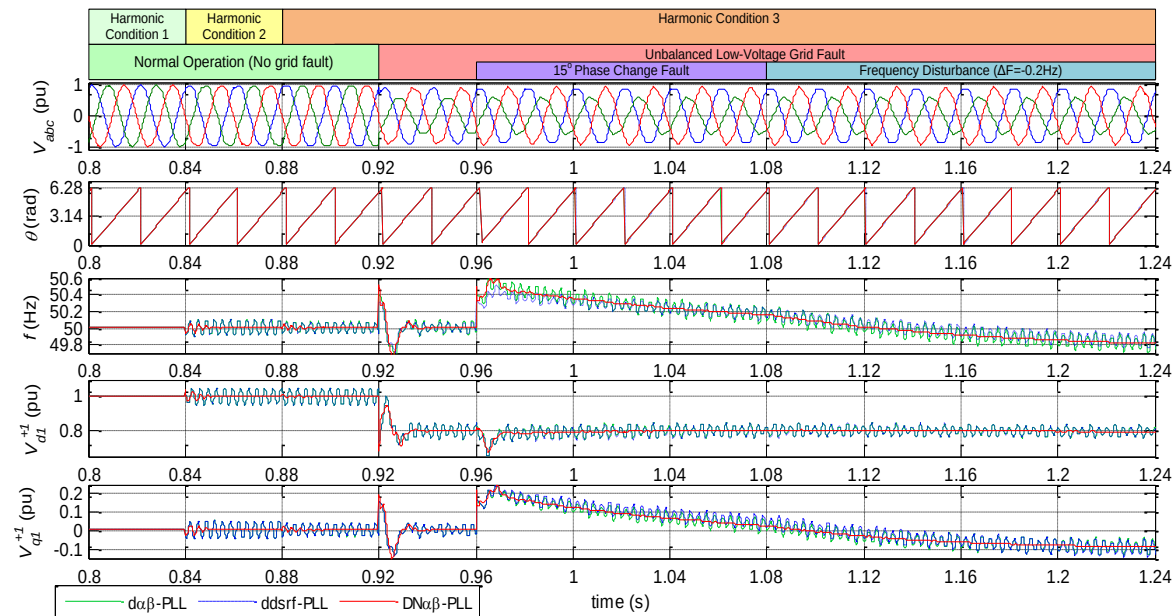
4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

DN $\alpha\beta$ -PLL (DN in $\alpha\beta$ -frame PLL)



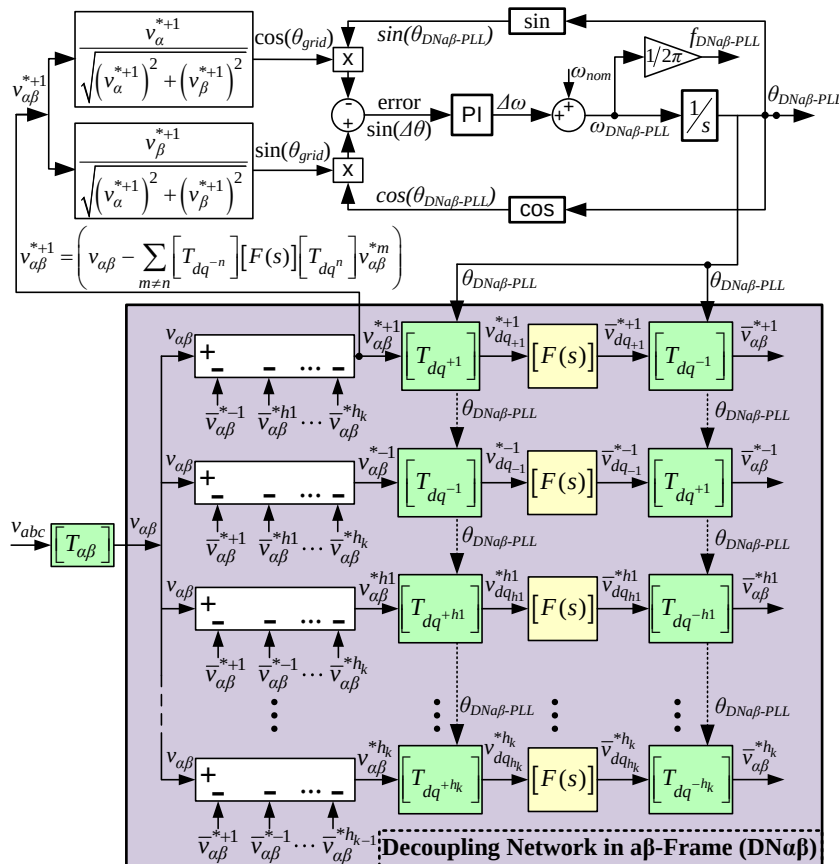
Structure of the DN $\alpha\beta$ -PLL



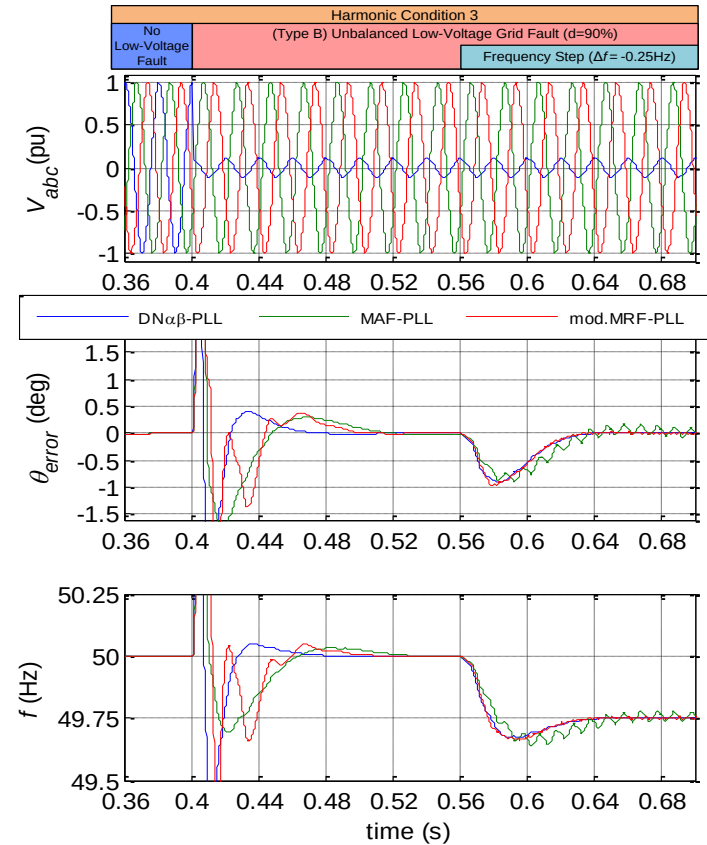
4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

DN $\alpha\beta$ -PLL (DN in $\alpha\beta$ -frame PLL)



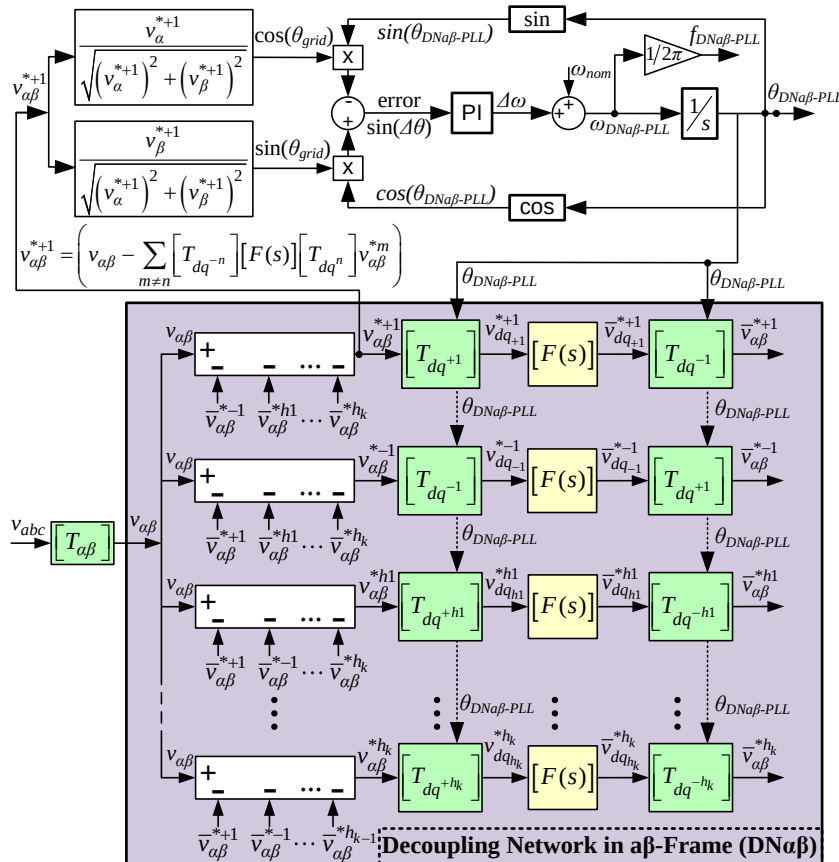
Structure of the DN $\alpha\beta$ -PLL



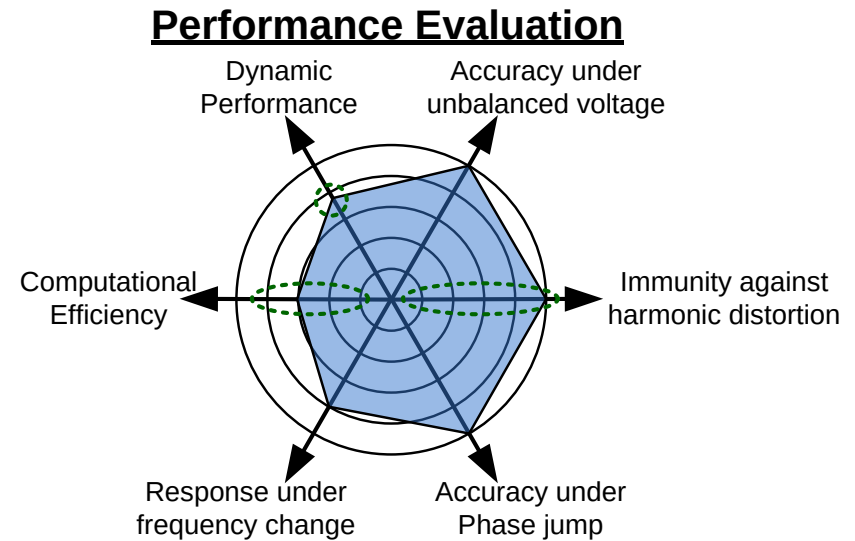
4. Synchronization Methods

Three-phase synchronization method: Conventional methods – State of the art – **Proposed methods**

DN $\alpha\beta$ -PLL (DN in $\alpha\beta$ -frame PLL)



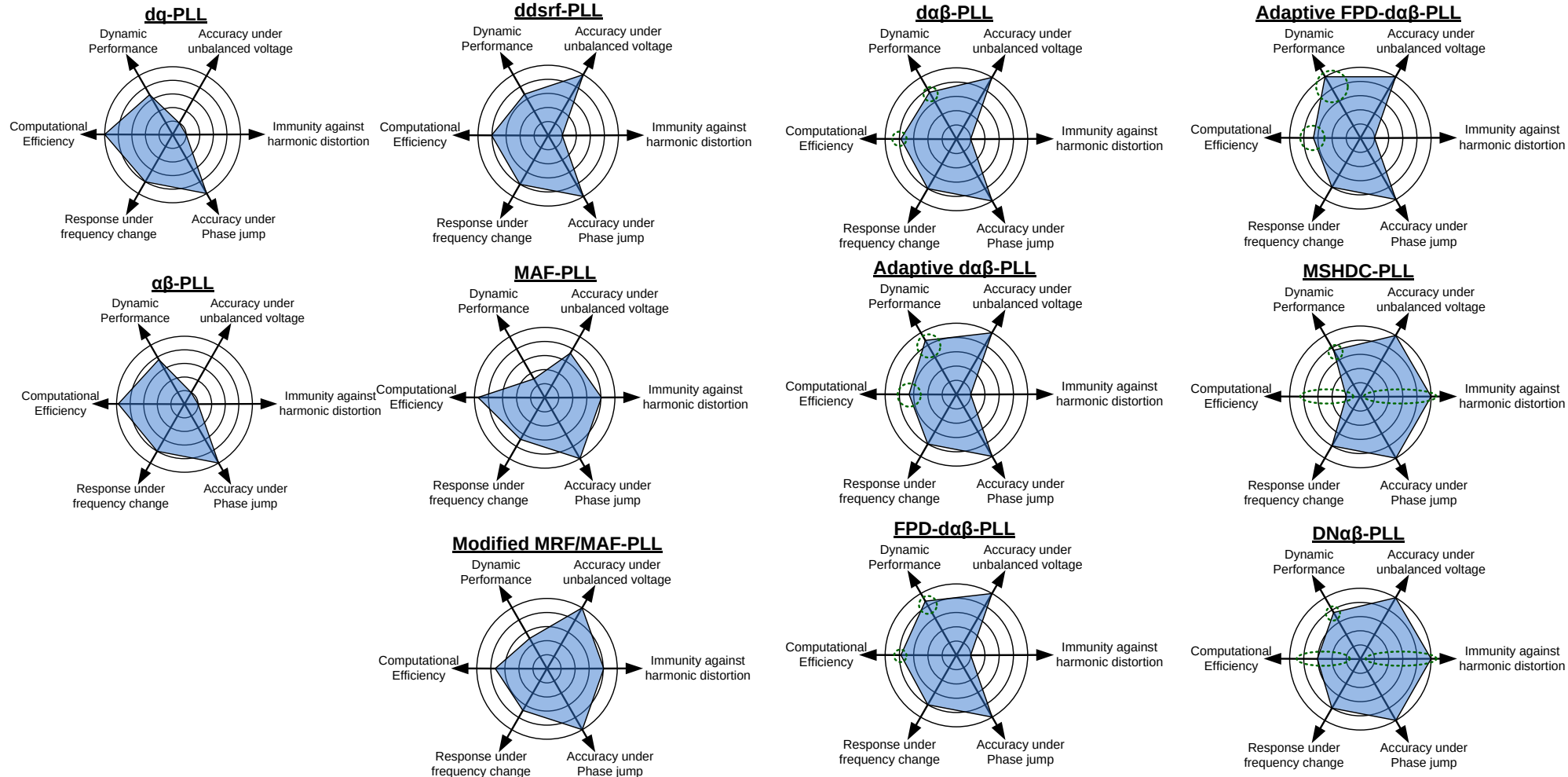
Structure of the DN $\alpha\beta$ -PLL





4. Synchronization Methods

Synopsis





5. Conclusions

Conclusions

This work has presented the control methodologies for the GSC of RES (WPS and PV systems)

- The main topologies for WPS and PV systems
- The main units of the GSC controller has been presented

Proposed current controller

- Simplified and can enable the accurate current injection under unbalanced and harmonic conditions

Literature review on the synchronization methods

- Conventional and state-of-the-art PLLs

Six new synchronization methods has been proposed

- The proposed PLLs enhance the synchronization in regards:
 - the dynamic performance
 - the immunity against unbalanced voltage
 - The immunity against harmonic distortion



5. Conclusions

Future Work

- Propose new synchronization methods for single-phase systems
- Propose new advanced current controller for both single- and three-phase systems
- Enhance PQ controller for both single- and three-phase systems
- Investigate the effect of large RES on the dynamic operation of power systems
- Investigate novel control strategies for PV and their effect on distribution networks



Acknowledgements

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Thank you for your attention

